

FEATURES

- Ultralow power:** as low as 23 μA in measurement mode and 0.2 μA in standby mode at $V_S = 2.6\text{ V}$ (typical)
- Power consumption scales automatically with bandwidth**
- User-selectable resolution**
 - Fixed 10-bit resolution
 - Full resolution, where resolution increases with g range, up to 13-bit resolution at $\pm 16\text{ g}$ (maintaining 4 mg/LSB scale factor in all g ranges)
- Embedded memory management system with FIFO technology minimizes host processor load**
- Single tap/double tap detection**
- Activity/inactivity monitoring**
- Free-fall detection**
- Concurrent four- and six-position orientation detection**
- Supply and I/O voltage range:** 1.7 V to 2.75 V
- SPI (3- and 4-wire) and I²C digital interfaces**
- Flexible interrupt modes mappable to either interrupt pin**
- Measurement ranges selectable via serial command**
- Bandwidth selectable via serial command**
- Wide temperature range** (-40°C to $+85^\circ\text{C}$)
- 10,000 g shock survival**
- Pb free/RoHS compliant**
- Small and thin:** 3 mm \times 3 mm \times 0.95 mm LGA package

APPLICATIONS

- Handsets
- Medical instrumentation
- Gaming and pointing devices
- Industrial instrumentation
- Personal navigation devices
- Hard disk drive (HDD) protection

GENERAL DESCRIPTION

The ADXL346 is a small, thin, ultralow power, 3-axis accelerometer with high resolution (13-bit) measurement at up to $\pm 16\text{ g}$. Digital output data is formatted as 16-bit twos complement and is accessible through either an SPI (3- or 4-wire) or I²C digital interface.

The ADXL346 is well suited for mobile device applications. It measures the static acceleration of gravity in tilt-sensing applications, as well as dynamic acceleration resulting from motion or shock. Its high resolution (4 mg/LSB) enables measurement of inclination changes of less than 1.0° .

Several special sensing functions are provided. Activity and inactivity sensing detect the presence or lack of motion by comparing the acceleration on any axis with user-set thresholds. Tap sensing detects single and double taps in any direction. Free-fall sensing detects if the device is falling. Orientation detection is capable of concurrent four- and six-position sensing and a user-selectable interrupt on orientation change for 2D or 3D applications. These functions can be mapped individually to either of two interrupt output pins. An integrated memory management system with 32-level first in, first out (FIFO) buffer can be used to store data to minimize host processor activity and lower overall system power consumption.

Low power modes enable intelligent motion-based power management with threshold sensing and active acceleration measurement at extremely low power dissipation.

The ADXL346 is supplied in a small, thin, 3 mm \times 3 mm \times 0.95 mm, 16-lead, plastic package.

FUNCTIONAL BLOCK DIAGRAM

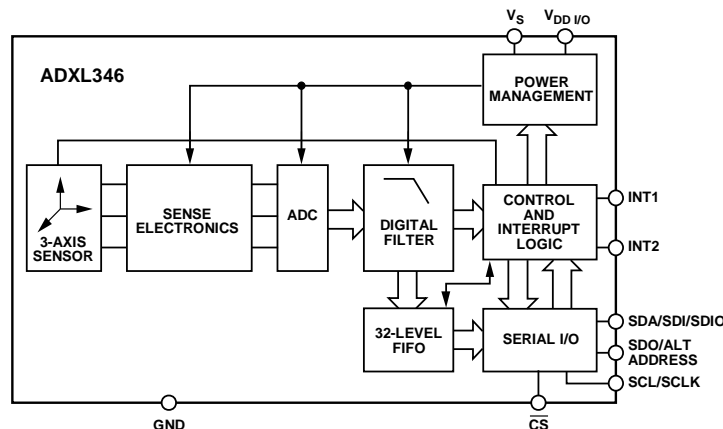


Figure 1.

Rev. C

[Document Feedback](#)

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ADXL346* PRODUCT PAGE QUICK LINKS

Last Content Update: 02/23/2017

COMPARABLE PARTS

View a parametric search of comparable parts.

EVALUATION KITS

- ADXL346 Datalogger / Development Board
- ADXL346Z Breakout Board
- Real Time Eval System for Digital Output Sensor

DOCUMENTATION

Application Notes

- AN-1025: Utilization of the First In, First Out (FIFO) Buffer in Analog Devices, Inc. Digital Accelerometers
- AN-1057: Using an Accelerometer for Inclination Sensing
- AN-1063: Oversampling Technique to Improve ADXL345 Output Resolution
- AN-1078: ADXL346 Quick Start Guide
- AN-688: Phase and Frequency Response of iMEMS® Accelerometers and Gyros

Data Sheet

- ADXL346: 3-Axis, ± 2 g/ ± 4 g/ ± 8 g/ ± 16 g Ultralow Power Digital Accelerometer Data Sheet

User Guides

- UG-012: μ MEMS ADXL346 Inertial Sensor Evaluation System User Guide

SOFTWARE AND SYSTEMS REQUIREMENTS

- ADXL345 Input 3-Axis Digital Accelerometer Linux Driver

TOOLS AND SIMULATIONS

- ADXL345/6 Accelerometer FIFO Over-sampling Calculator

DESIGN RESOURCES

- ADXL346 Material Declaration
- PCN-PDN Information
- Quality And Reliability
- Symbols and Footprints

DISCUSSIONS

View all ADXL346 EngineerZone Discussions.

SAMPLE AND BUY

Visit the product page to see pricing options.

TECHNICAL SUPPORT

Submit a technical question or find your regional support number.

DOCUMENT FEEDBACK

Submit feedback for this data sheet.

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REVISION HISTORY

6/15—Rev. B to Rev. C

Changes to Features Section and General Description Section	1
Change to Figure 36	14
Change to FIFO Section	20

5/11—Rev. A to Rev. B

Added Endnote 7	3
Added Preventing Bus Traffic Errors Section	14
Changes to Figure 37, Figure 38, Figure 39	15
Changes to Table 12	18
Changes to Using Self-Test Section	32
Changes to Axes of Acceleration Sensitivity Section	37

11/10—Rev. 0 to Rev. A

Changes to Ordering Guide	39
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5/10—Revision 0: Initial Version

SPECIFICATIONS

$T_A = 25^\circ\text{C}$, $V_S = 2.6\text{ V}$, $V_{DD\text{ I/O}} = 1.8\text{ V}$, acceleration = 0 g, $C_S = 10\text{ }\mu\text{F}$ tantalum, $C_{\text{I/O}} = 0.1\text{ }\mu\text{F}$, ODR = 800 Hz, unless otherwise noted.

Table 1. Specifications

Parameter	Test Conditions	Min ¹	Typ ²	Max ¹	Unit
SENSOR INPUT					
Measurement Range	Each axis User selectable		$\pm 2, \pm 4, \pm 8, \pm 16$		g
Nonlinearity	Percentage of full scale		± 0.5		%
Inter-Axis Alignment Error			± 0.1		Degrees
Cross-Axis Sensitivity ³			± 1		%
OUTPUT RESOLUTION					
All g Ranges	Each axis 10-bit resolution		10		Bits
$\pm 2\text{ g}$ Range	Full resolution		10		Bits
$\pm 4\text{ g}$ Range	Full resolution		11		Bits
$\pm 8\text{ g}$ Range	Full resolution		12		Bits
$\pm 16\text{ g}$ Range	Full resolution		13		Bits
SENSITIVITY					
Sensitivity at $X_{\text{OUT}}, Y_{\text{OUT}}, Z_{\text{OUT}}$	Each axis All g ranges, full resolution	230	256	282	LSB/g
	$\pm 2\text{ g}$, 10-bit resolution	230	256	282	LSB/g
	$\pm 4\text{ g}$, 10-bit resolution	115	128	141	LSB/g
	$\pm 8\text{ g}$, 10-bit resolution	57	64	71	LSB/g
	$\pm 16\text{ g}$, 10-bit resolution	29	32	35	LSB/g
Sensitivity Deviation from Ideal Scale Factor at $X_{\text{OUT}}, Y_{\text{OUT}}, Z_{\text{OUT}}$	All g ranges		± 1.0		%
	All g ranges, full resolution	3.5	3.9	4.3	mg/LSB
	$\pm 2\text{ g}$, 10-bit resolution	3.5	3.9	4.3	mg/LSB
	$\pm 4\text{ g}$, 10-bit resolution	7.1	7.8	8.7	mg/LSB
	$\pm 8\text{ g}$, 10-bit resolution	14.1	15.6	17.5	mg/LSB
Sensitivity Change Due to Temperature	$\pm 16\text{ g}$, 10-bit resolution	28.6	31.2	34.5	mg/LSB
			± 0.02		%/ $^\circ\text{C}$
0 g OFFSET					
0 g Output for $X_{\text{OUT}}, Y_{\text{OUT}}, Z_{\text{OUT}}$	Each axis	-150	0	+150	mg
0 g Output Deviation from Ideal			± 35		mg
0 g Offset vs. Temperature for X-, Y-Axes			± 0.7		mg/ $^\circ\text{C}$
0 g Offset vs. Temperature for Z-Axis			± 1.3		mg/ $^\circ\text{C}$
NOISE					
X-, Y-Axes	ODR = 100 Hz for $\pm 2\text{ g}$, 10-bit resolution or all g ranges, full resolution		1.1		LSB rms
Z-Axis	ODR = 100 Hz for $\pm 2\text{ g}$, 10-bit resolution or all g ranges, full resolution		1.5		LSB rms
OUTPUT DATA RATE AND BANDWIDTH					
Output Data Rate (ODR) ^{4, 5, 6, 7}	User selectable	0.10		3200	Hz
SELF-TEST⁸					
Output Change in X-Axis		0.27		1.55	g
Output Change in Y-Axis		-1.55		-0.27	g
Output Change in Z-Axis		0.40		1.95	g
POWER SUPPLY					
Operating Voltage Range (V_S)		1.7	2.6	2.75	V
Interface Voltage Range ($V_{DD\text{ I/O}}$)		1.7	1.8	V_S	V
Measurement Mode Supply Current	ODR $\geq 100\text{ Hz}$		140		μA
	ODR $< 10\text{ Hz}$		30		μA
Standby Mode Supply Current			0.2		μA
Turn-On and Wake-Up Time ⁹	ODR = 3200 Hz		1.4		ms

Parameter	Test Conditions	Min ¹	Typ ²	Max ¹	Unit
TEMPERATURE Operating Temperature Range		-40		+85	°C
WEIGHT Device Weight			18		mg

¹ All minimum and maximum specifications are guaranteed. Typical specifications are not guaranteed.

² The typical specifications shown are for at least 68% of the population of parts and are based on the worst case of mean $\pm 1 \sigma$ except for 0 g output and sensitivity, which represents the target value. For 0 g offset and sensitivity, the deviation from the ideal describes the worst case of mean $\pm 1 \sigma$.

³ Cross-axis sensitivity is defined as coupling between any two axes.

⁴ Bandwidth is the -3 dB frequency and is half the output data rate bandwidth = ODR/2.

⁵ The output format for the 3200 Hz and 1600 Hz ODRs is different from the output format for the remaining ODRs. This difference is described in the Data Formatting of Upper Data Rates section.

⁶ Output data rates below 6.25 Hz exhibit additional offset shift with increased temperature, depending on selected output data rate. Refer to the Offset Performance at Lowest Data Rates section for details.

⁷ These are typical values for the lowest and highest output data rate settings.

⁸ Self-test change is defined as the output (g) when the SELF_TEST bit = 1 (in the DATA_FORMAT register, Address 0x31) minus the output (g) when the SELF_TEST bit = 0. Due to device filtering, the output reaches its final value after $4 \times \tau$ when enabling or disabling self-test, where $\tau = 1/(\text{data rate})$. The part must be in normal power operation (LOW_POWER bit = 0 in the BW_RATE register, Address 0x2C) for self-test to operate correctly.

⁹ Turn-on and wake-up times are determined by the user-defined bandwidth. At a 100 Hz data rate, the turn-on and wake-up times are each approximately 11.1 ms. For other data rates, the turn-on and wake-up times are each approximately $\tau + 1.1$ in milliseconds, where $\tau = 1/(\text{data rate})$.

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Acceleration	
Any Axis, Unpowered	10,000 g
Any Axis, Powered	10,000 g
V_S	-0.3 V to +3.0 V
$V_{DD I/O}$	-0.3 V to +3.0 V
Digital Pins	-0.3 V to $V_{DD I/O} + 0.3$ V or 3.0 V, whichever is less
All Other Pins	-0.3 V to +3.0 V
Output Short-Circuit Duration (Any Pin to Ground)	Indefinite
Temperature Range	
Powered	-40°C to +105°C
Storage	-40°C to +105°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Table 3. Package Characteristics

Package Type	θ_{JA}	θ_{JC}	Device Weight
16-Terminal LGA	150°C/W	85°C/W	18 mg

PACKAGE INFORMATION

The information in Figure 2 and Table 4 provide details about the package branding for [ADXL346](#). For a complete listing of product availability, see the Ordering Guide section.

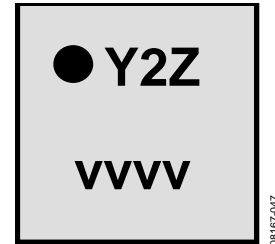


Figure 2. Product Information on Package (Top View)

Table 4. Package Branding Information

Branding Key	Field Description
Y2Z	Part identifier for ADXL346
vvvv	Factory lot code

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

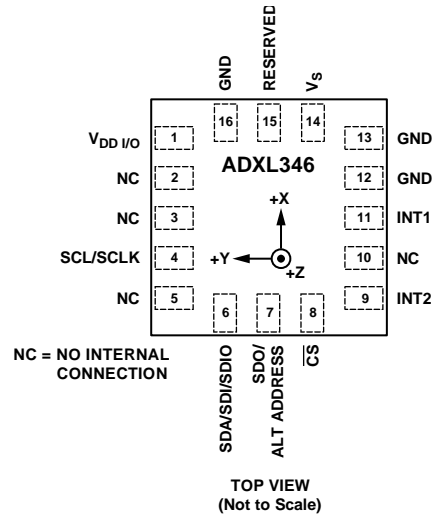


Figure 3. Pin Configuration (Top View)

Table 5. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	V _{DD I/O}	Digital Interface Supply Voltage.
2	NC	Not Internally Connected.
3	NC	Not Internally Connected.
4	SCL/SCLK	Serial Communications Clock.
5	NC	Not Internally Connected.
6	SDA/SDI/SDIO	Serial Data (I ² C)/Serial Data Input (SPI 4-Wire)/Serial Data Input and Output (SPI 3-Wire).
7	SDO/ALT ADDRESS	Serial Data Output (SPI 4-Wire)/Alternate I ² C Address Select (I ² C).
8	\overline{CS}	Chip Select.
9	INT2	Interrupt 2 Output.
10	NC	Not Internally Connected.
11	INT1	Interrupt 1 Output.
12	GND	Must be connected to ground.
13	GND	Must be connected to ground.
14	V _s	Supply Voltage.
15	RESERVED	Reserved. This pin must be connected to V _s .
16	GND	Must be connected to ground.

TYPICAL PERFORMANCE CHARACTERISTICS

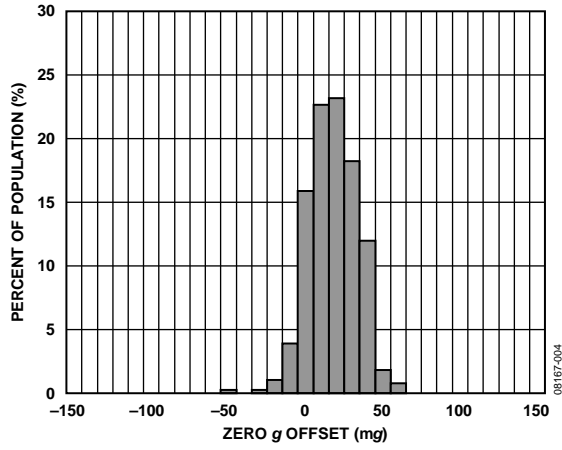


Figure 4. X-Axis Zero g Offset at 25°C, $V_S = 2.6 V$

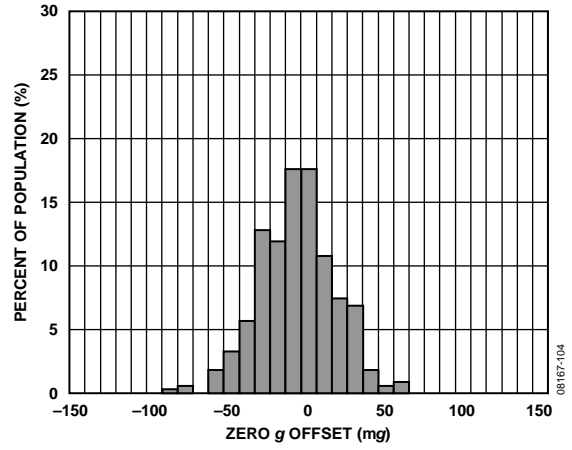


Figure 7. X-Axis Zero g Offset at 25°C, $V_S = 1.8 V$

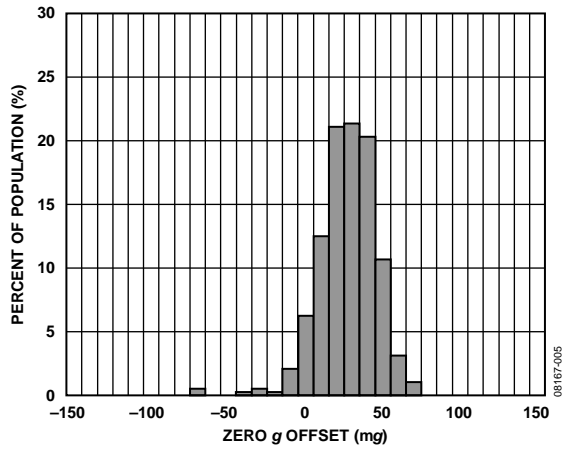


Figure 5. Y-Axis Zero g Offset at 25°C, $V_S = 2.6 V$

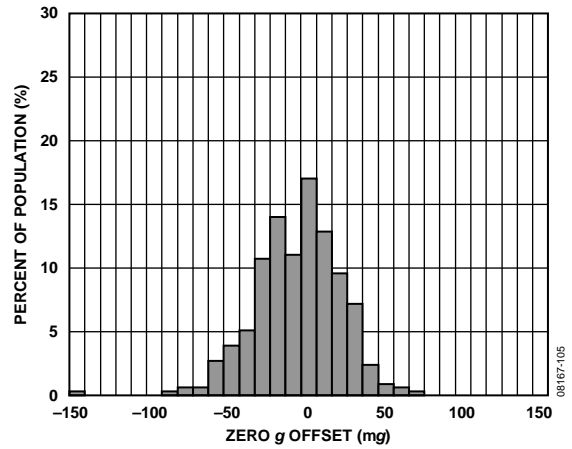


Figure 8. Y-Axis Zero g Offset at 25°C, $V_S = 1.8 V$

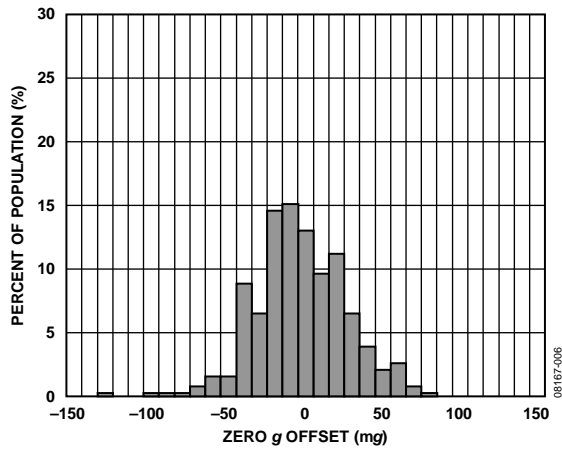


Figure 6. Z-Axis Zero g Offset at 25°C, $V_S = 2.6 V$

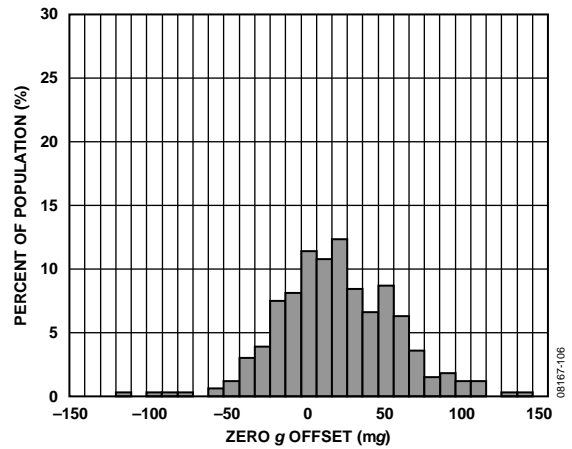


Figure 9. Z-Axis Zero g Offset at 25°C, $V_S = 1.8 V$

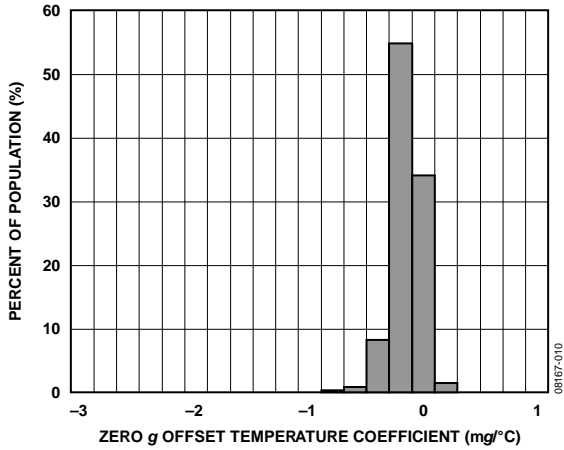


Figure 10. X-Axis Zero g Offset Temperature Coefficient, $V_s = 2.6 V$

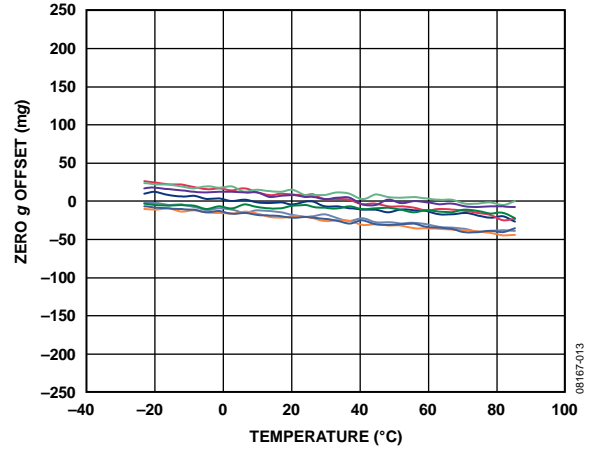


Figure 13. X-Axis Zero g Offset vs. Temperature—
Eight Parts Soldered to PCB, $V_s = 2.6 V$

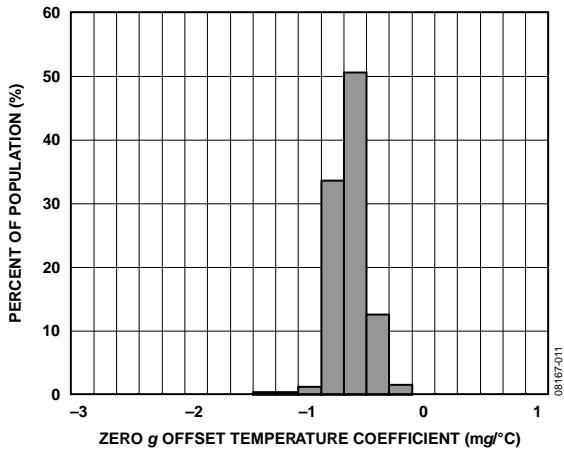


Figure 11. Y-Axis Zero g Offset Temperature Coefficient, $V_s = 2.6 V$

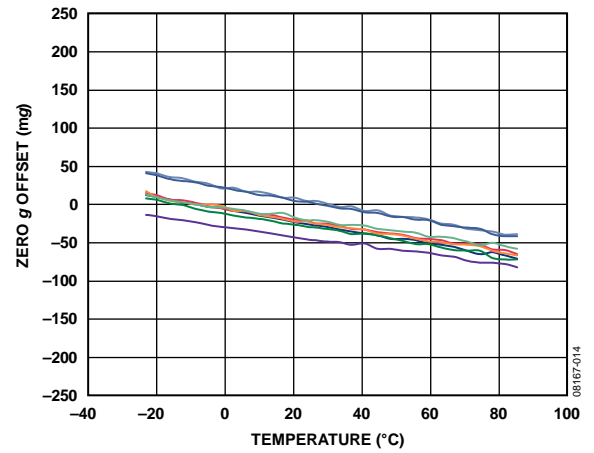


Figure 14. Y-Axis Zero g Offset vs. Temperature—
Eight Parts Soldered to PCB, $V_s = 2.6 V$

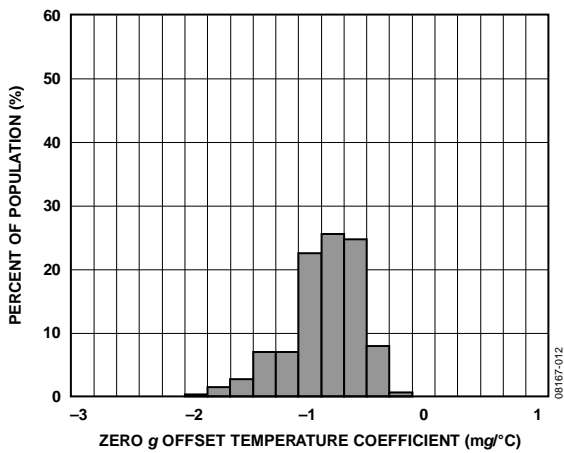


Figure 12. Z-Axis Zero g Offset Temperature Coefficient, $V_s = 2.6 V$

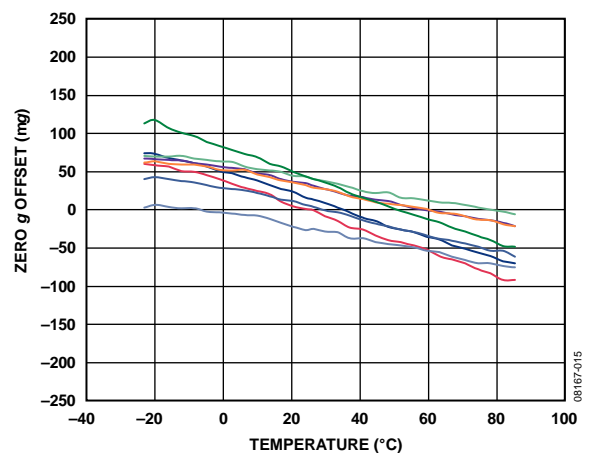


Figure 15. Z-Axis Zero g Offset vs. Temperature—
Eight Parts Soldered to PCB, $V_s = 2.6 V$

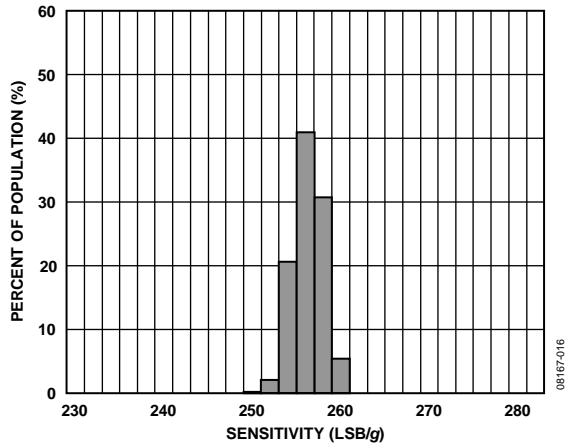


Figure 16. X-Axis Sensitivity at 25°C, Vs = 2.6 V, Full Resolution

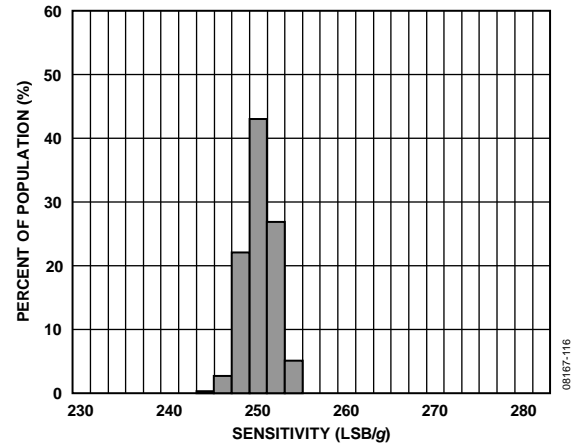


Figure 19. X-Axis Sensitivity at 25°C, Vs = 1.8 V, Full Resolution

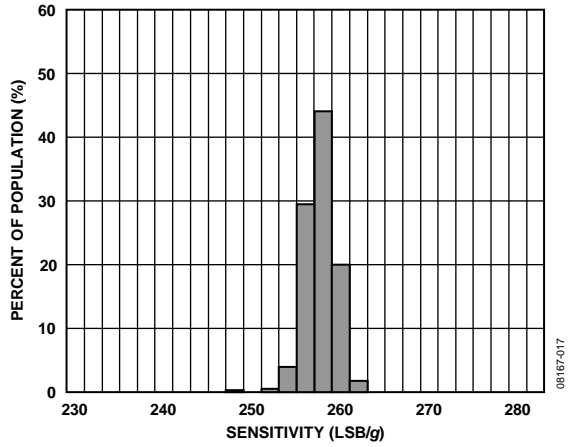


Figure 17. Y-Axis Sensitivity at 25°C, Vs = 2.6 V, Full Resolution

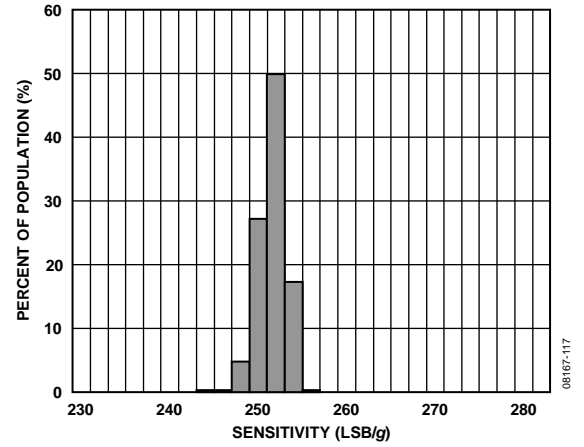


Figure 20. Y-Axis Sensitivity at 25°C, Vs = 1.8 V, Full Resolution

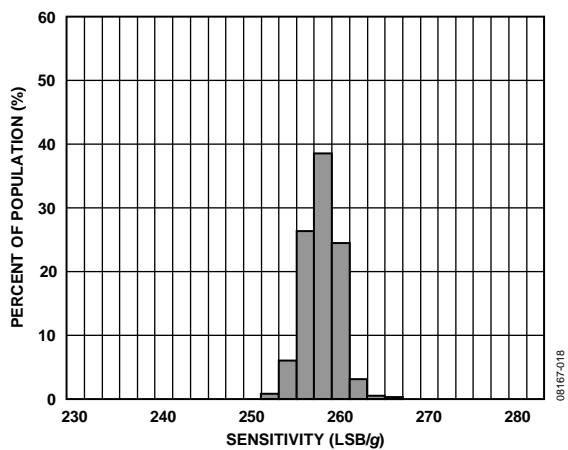


Figure 18. Z-Axis Sensitivity at 25°C, Vs = 2.6 V, Full Resolution

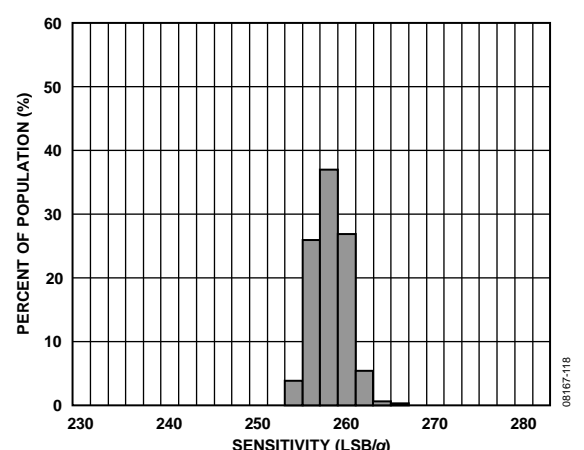


Figure 21. Z-Axis Sensitivity at 25°C, Vs = 1.8 V, Full Resolution

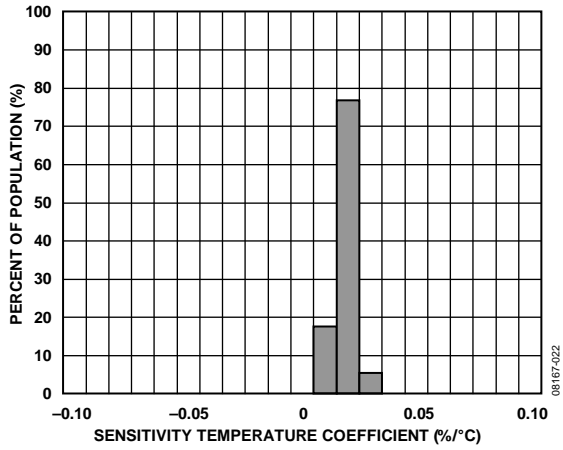


Figure 22. X-Axis Sensitivity Temperature Coefficient, $V_S = 2.6\text{ V}$

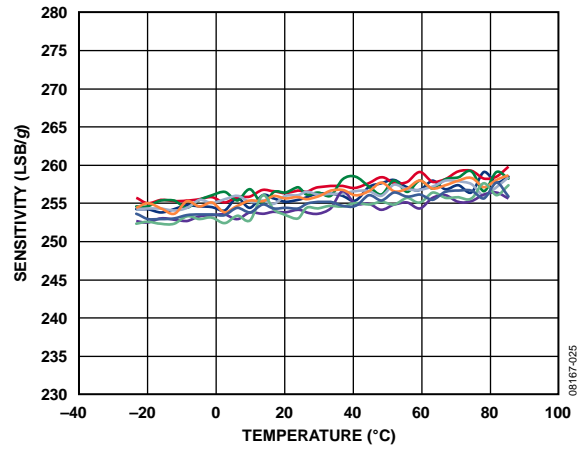


Figure 25. X-Axis Sensitivity vs. Temperature—
Eight Parts Soldered to PCB, $V_S = 2.6\text{ V}$, Full Resolution

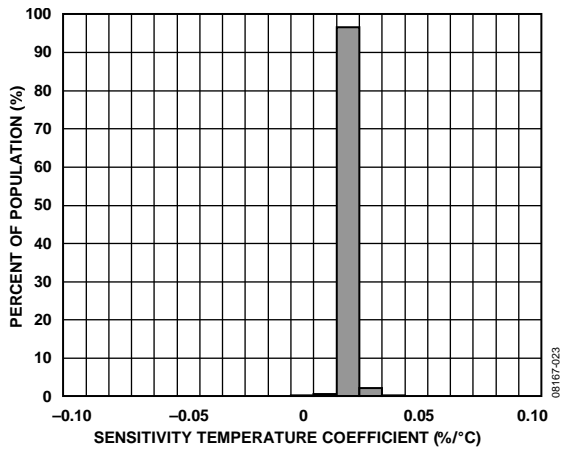


Figure 23. Y-Axis Sensitivity Temperature Coefficient, $V_S = 2.6\text{ V}$

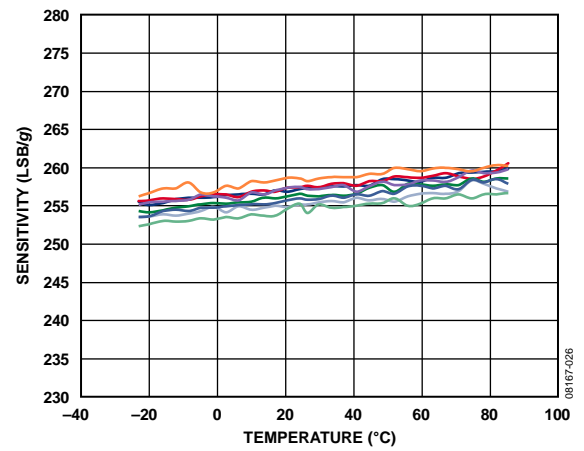


Figure 26. Y-Axis Sensitivity vs. Temperature—
Eight Parts Soldered to PCB, $V_S = 2.6\text{ V}$, Full Resolution

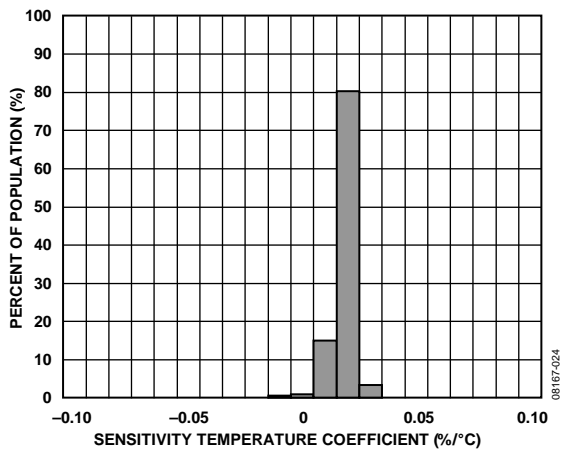


Figure 24. Z-Axis Sensitivity Temperature Coefficient, $V_S = 2.6\text{ V}$

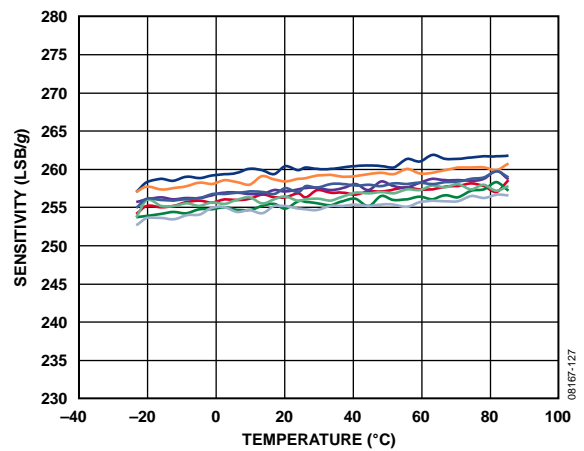


Figure 27. Z-Axis Sensitivity vs. Temperature—
Eight Parts Soldered to PCB, $V_S = 2.6\text{ V}$, Full Resolution

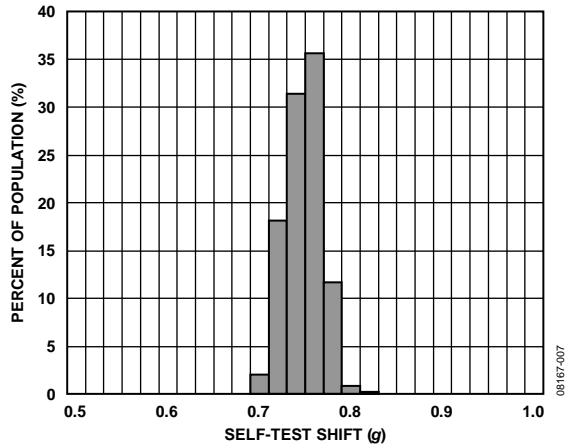


Figure 28. X-Axis Self-Test Response at 25°C, $V_S = 2.6 V$

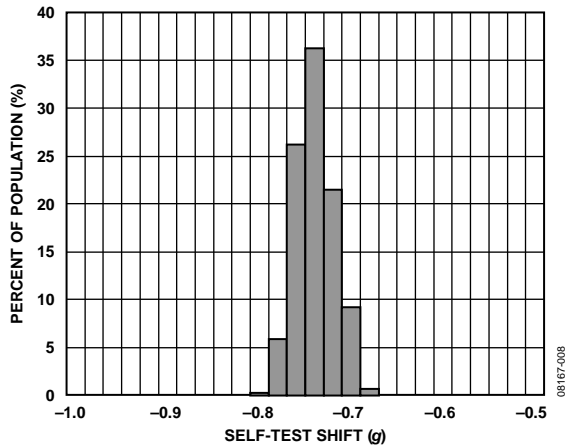


Figure 29. Y-Axis Self-Test Response at 25°C, $V_S = 2.6 V$

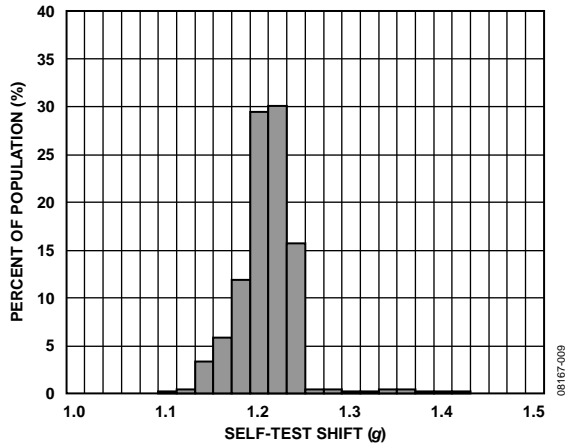


Figure 30. Z-Axis Self-Test Response at 25°C, $V_S = 2.6 V$

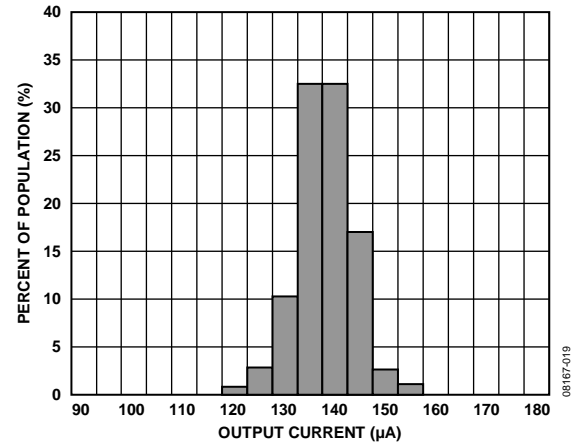


Figure 31. Supply Current at 25°C, 100 Hz Output Data Rate, $V_S = 2.6 V$

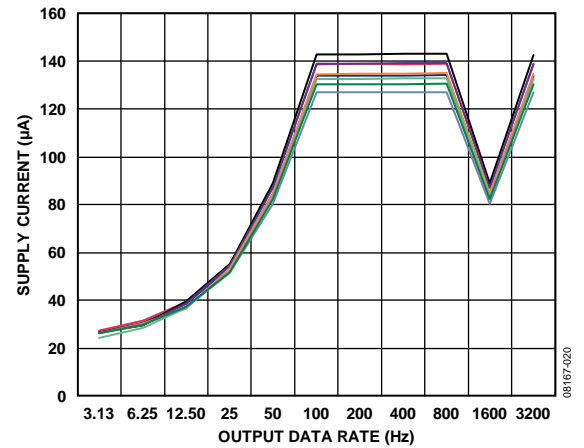


Figure 32. Supply Current vs. Output Data Rate at 25°C—10 Parts, $V_S = 2.6 V$

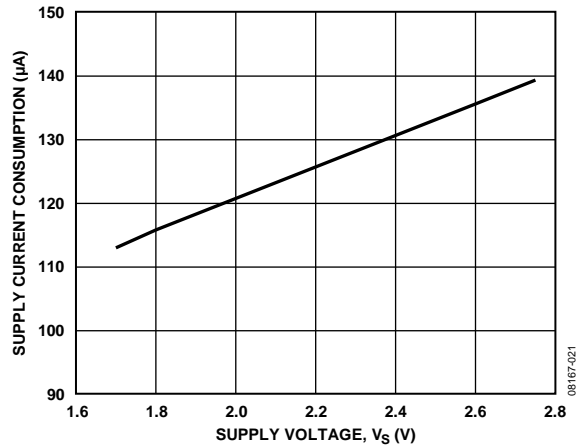


Figure 33. Supply Current vs. Supply Voltage at 25°C

THEORY OF OPERATION

The ADXL346 is a complete 3-axis acceleration measurement system with a selectable measurement range of $\pm 2\text{ g}$, $\pm 4\text{ g}$, $\pm 8\text{ g}$, or $\pm 16\text{ g}$. It measures both dynamic acceleration resulting from motion or shock and static acceleration, such as gravity, which allows the device to be used as a tilt sensor.

The sensor is a polysilicon surface-micromachined structure built on top of a silicon wafer. Polysilicon springs suspend the structure over the surface of the wafer and provide a resistance against forces due to applied acceleration.

Deflection of the structure is measured using differential capacitors that consist of independent fixed plates and plates attached to the moving mass. Acceleration deflects the proof mass and unbalances the differential capacitor, resulting in a sensor output with an amplitude proportional to acceleration. Phase-sensitive demodulation is used to determine the magnitude and polarity of the acceleration.

POWER SEQUENCING

Power can be applied to V_S or $V_{DD I/O}$ in any sequence without damaging the ADXL346. All possible power-on modes are summarized in Table 6. The interface voltage level is set with the interface supply voltage, $V_{DD I/O}$, which must be present to ensure that the ADXL346 does not create a conflict on the communication bus. For single-supply operation, $V_{DD I/O}$ can be the same as the main supply, V_S . In a dual-supply application, however, $V_{DD I/O}$ can differ from V_S to accommodate the desired interface voltage, as long as V_S is greater than or equal to $V_{DD I/O}$.

After V_S is applied, the device enters standby mode, where power consumption is minimized and the device waits for $V_{DD I/O}$ to be applied and for the command to enter measurement mode to be received. (This command can be initiated by setting the measure bit (Bit D3) in the POWER_CTL register (Address 0x2D).) In addition, any register can be written to or read from to configure the part while the device is in standby mode. It is recommended to configure the device in standby mode and then to enable measurement mode. Clearing the measure bit returns the device to the standby mode.

Table 6. Power Sequencing

Condition	V_S	$V_{DD I/O}$	Description
Power Off	Off	Off	The device is completely off, but there is a potential for a communication bus conflict.
Bus Disabled	On	Off	The device is on in standby mode, but communication is unavailable and will create a conflict on the communication bus. The duration of this state should be minimized during power-up to prevent a conflict.
Bus Enabled	Off	On	No functions are available, but the device will not create a conflict on the communication bus.
Standby or Measurement Mode	On	On	At power-up, the device is in standby mode, awaiting a command to enter measurement mode, and all sensor functions are off. After the device is instructed to enter measurement mode, all sensor functions are available.

POWER SAVINGS

Power Modes

The ADXL346 automatically modulates its power consumption in proportion to its output data rate, as outlined in Table 7. If additional power savings is desired, a lower power mode is available. In this mode, the internal sampling rate is reduced, allowing for power savings in the 12.5 Hz to 400 Hz data rate range at the expense of slightly greater noise. To enter low power mode, set the LOW_POWER bit (Bit D4) in the BW_RATE register (Address 0x2C). The current consumption in low power mode is shown in Table 8 for cases where there is an advantage to using low power mode. Use of low power mode for a data rate not shown in Table 8 does not provide any advantage over the same data rate in normal power mode. Therefore, it is recommended that only data rates listed in Table 8 be used in low power mode. The current consumption values shown in Table 7 and Table 8 are for a V_S of 2.6 V.

Table 7. Typical Current Consumption vs. Data Rate
($T_A = 25^\circ\text{C}$, $V_S = 2.6\text{ V}$, $V_{DD I/O} = 1.8\text{ V}$)

Output Data Rate (Hz)	Bandwidth (Hz)	Rate Code	I_{DD} (μA)
3200	1600	1111	140
1600	800	1110	90
800	400	1101	140
400	200	1100	140
200	100	1011	140
100	50	1010	140
50	25	1001	90
25	12.5	1000	55
12.5	6.25	0111	40
6.25	3.13	0110	31
3.13	1.56	0101	27
1.56	0.78	0100	23
0.78	0.39	0011	23
0.39	0.20	0010	23
0.20	0.10	0001	23
0.10	0.05	0000	23

Table 8. Typical Current Consumption vs. Data Rate, Low Power Mode ($T_A = 25^\circ\text{C}$, $V_S = 2.6\text{ V}$, $V_{DD I/O} = 1.8\text{ V}$)

Output Data Rate (Hz)	Bandwidth (Hz)	Rate Code	I_{DD} (μA)
400	200	1100	90
200	100	1011	55
100	50	1010	40
50	25	1001	31
25	12.5	1000	27
12.5	6.25	0111	23

Autosleep Mode

Additional power can be saved if the ADXL346 automatically switches to sleep mode during periods of inactivity. To enable this feature, set the THRESH_INACT register (Address 0x25) and the TIME_INACT register (Address 0x26) each to a value that signifies inactivity (the appropriate value depends on the application), and then set the AUTO_SLEEP bit (Bit D4) and the link bit (Bit D5) in the POWER_CTL register (Address 0x2D). Current consumption at the sub-8 Hz data rates used in this mode is typically 23 μA for a V_S of 2.6 V.

Standby Mode

For even lower power operation, standby mode can be used. In standby mode, current consumption is reduced to 0.2 μA (typical). In this mode, no measurements are made. Standby mode is entered by clearing the measure bit (Bit D3) in the POWER_CTL register (Address 0x2D). Placing the device into standby mode preserves the contents of FIFO.

SERIAL COMMUNICATIONS

I²C and SPI digital communications are available. In both cases, the ADXL346 operates as a slave. I²C mode is enabled if the \overline{CS} pin is tied high to $V_{DD I/O}$. The \overline{CS} pin should always be tied high to $V_{DD I/O}$ or be driven by an external controller because there is no default mode if the \overline{CS} pin is left unconnected. Therefore, not taking these precautions may result in an inability to communicate with the part. In SPI mode, the \overline{CS} pin is controlled by the bus master. In both SPI and I²C modes of operation, data transmitted from the ADXL346 to the master device should be ignored during writes to the ADXL346.

SPI

For SPI, either 3- or 4-wire configuration is possible, as shown in the connection diagrams in Figure 34 and Figure 35. Clearing the SPI bit (Bit D6) in the DATA_FORMAT register (Address 0x31) selects 4-wire mode, whereas setting the SPI bit selects 3-wire mode. The maximum SPI clock speed is 5 MHz with 100 pF maximum loading, and the timing scheme follows clock polarity (CPOL) = 1 and clock phase (CPHA) = 1. If power is applied to the ADXL346 before the clock polarity and phase of the host processor are configured, the \overline{CS} pin should be brought high before changing the clock polarity and phase. When using 3-wire SPI, it is recommended that the SDO pin either be pulled up to $V_{DD I/O}$ or be pulled down to GND via a 10 k Ω resistor.

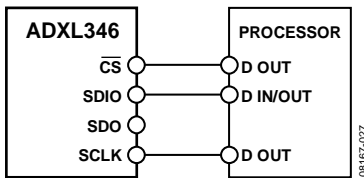


Figure 34. 3-Wire SPI Connection Diagram

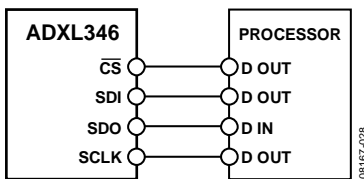


Figure 35. 4-Wire SPI Connection Diagram

\overline{CS} is the serial port enable line and is controlled by the SPI master. This line must go low at the start of a transmission and high at the end of a transmission, as shown in Figure 37. SCLK is the serial port clock and is supplied by the SPI master. SCLK should idle high during a period of no transmission. SDI and SDO are the serial data input and output, respectively. Data is updated on the falling edge of SCLK and should be sampled on the rising edge of SCLK.

To read or write multiple bytes in a single transmission, the multiple-byte bit, located after the R/W bit in the first byte transfer (MB in Figure 37 to Figure 39), must be set. After the register addressing and the first byte of data, each subsequent set of clock pulses (eight clock pulses) causes the ADXL346 to point to the next register for a read or write. This shifting continues until the clock pulses cease and \overline{CS} is deasserted. To perform reads or writes on different, nonsequential registers, \overline{CS} must be deasserted between transmissions and the new register must be addressed separately.

The timing diagram for 3-wire SPI reads or writes is shown in Figure 39. The 4-wire equivalents for SPI writes and reads are shown in Figure 37 and Figure 38, respectively. For correct operation of the part, the logic thresholds and timing parameters in Table 9 and Table 10 must be met at all times.

Use of the 3200 Hz and 1600 Hz output data rates is recommended only with SPI communication rates greater than or equal to 2 MHz. The 800 Hz output data rate is recommended only for communication speeds greater than or equal to 400 kHz, and the remaining data rates scale proportionally. For example, the minimum recommended communication speed for a 200 Hz output data rate is 100 kHz. Operation at an output data rate above the recommended maximum may result in undesirable effects on the acceleration data, including missing samples or additional noise.

Preventing Bus Traffic Errors

The ADXL346 \overline{CS} pin is used both for initiating SPI transactions, and for enabling I²C mode. When the ADXL346 is used on a SPI bus with multiple devices, its \overline{CS} pin is held high while the master communicates with the other devices. There may be conditions where a SPI command transmitted to another device looks like a valid I²C command. In this case, the ADXL346 would interpret this as an attempt to communicate in I²C mode, and could interfere with other bus traffic. Unless bus traffic can be adequately controlled to assure such a condition never occurs, it is recommended to add a logic gate in front of the SDI pin as shown in Figure 36. This OR gate will hold the SDA line high when \overline{CS} is high to prevent SPI bus traffic at the ADXL346 from appearing as an I²C start command.

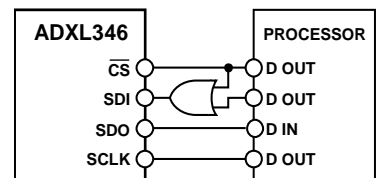


Figure 36. Recommended SPI Connection Diagram when Using Multiple SPI Devices on a Single Bus

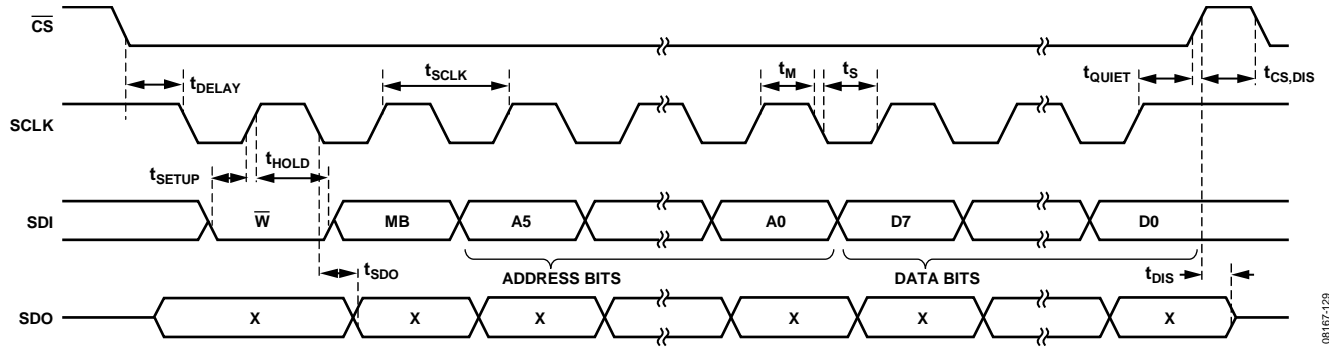


Figure 37. SPI 4-Wire Write

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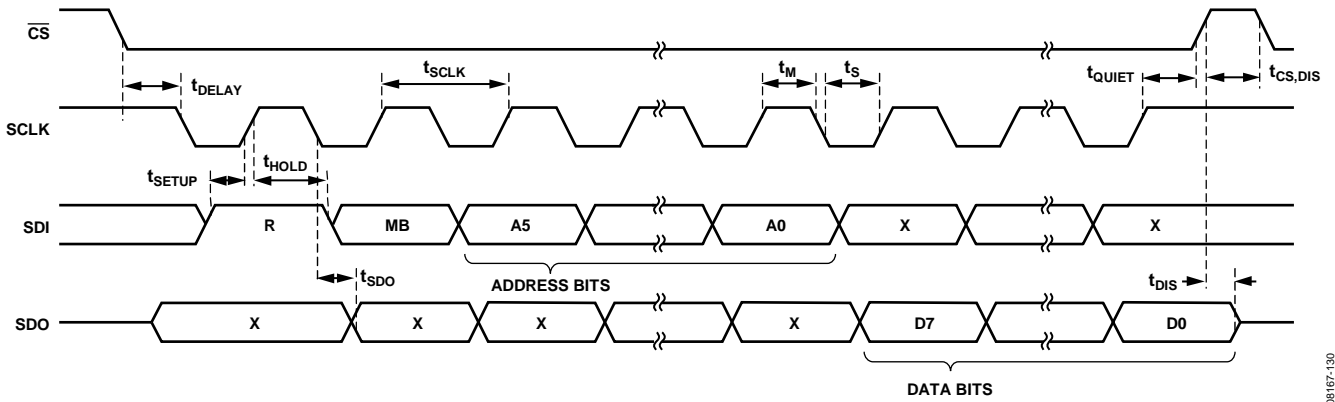
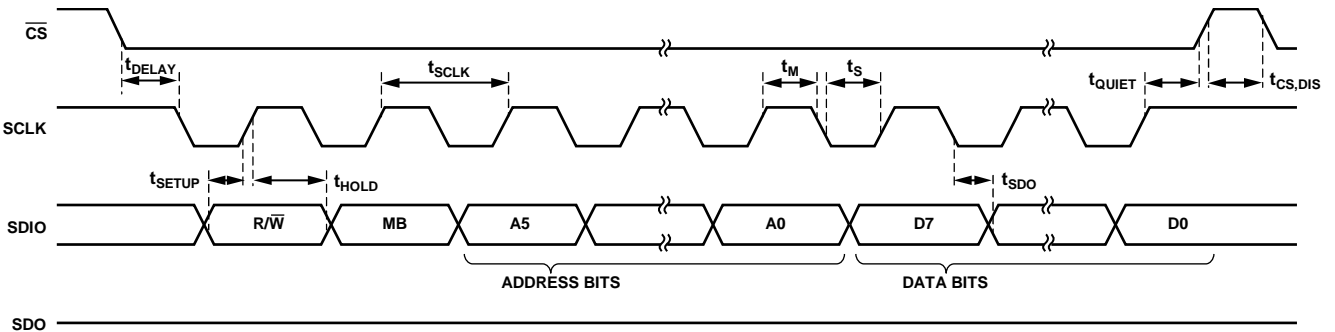


Figure 38. SPI 4-Wire Read

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NOTES

1. t_{SDO} IS ONLY PRESENT DURING READS.

Figure 39. SPI 3-Wire Read/Write

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Table 9. SPI Digital Input/Output

Parameter	Test Conditions	Limit ¹		Unit
		Min	Max	
Digital Input				
Low Level Input Voltage (V _{IL})			0.3 × V _{DD I/O}	V
High Level Input Voltage (V _{IH})		0.7 × V _{DD I/O}		V
Low Level Input Current (I _{IL})	V _{IN} = V _{DD I/O}		0.1	μA
High Level Input Current (I _{IH})	V _{IN} = 0 V	-0.1		μA
Digital Output				
Low Level Output Voltage (V _{OL})	I _{OL} = 10 mA		0.2 × V _{DD I/O}	V
High Level Output Voltage (V _{OH})	I _{OH} = -4 mA	0.8 × V _{DD I/O}		V
Low Level Output Current (I _{OL})	V _{OL} = V _{OL, max}	10		mA
High Level Output Current (I _{OH})	V _{OH} = V _{OH, min}		-4	mA
Pin Capacitance	f _{IN} = 1 MHz, V _{IN} = 2.6 V		8	pF

¹ Limits are based on characterization results; not production tested.

Table 10. SPI Timing (T_A = 25°C, V_S = 2.6 V, V_{DD I/O} = 1.8 V)¹

Parameter	Limit ^{2,3}		Unit	Description
	Min	Max		
f _{SCLK}		5	MHz	SPI clock frequency
t _{SCLK}	200		ns	1/(SPI clock frequency) mark-space ratio for the SCLK input is 40/60 to 60/40
t _{DELAY}	5		ns	\overline{CS} falling edge to SCLK falling edge
t _{QUIET}	5		ns	SCLK rising edge to \overline{CS} rising edge
t _{DIS}		10	ns	\overline{CS} rising edge to SDO disabled
t _{\overline{CS},DIS}	150		ns	\overline{CS} deassertion between SPI communications
t _S	0.3 × t _{SCLK}		ns	SCLK low pulse width (space)
t _M	0.3 × t _{SCLK}		ns	SCLK high pulse width (mark)
t _{SETUP}	5		ns	SDI valid before SCLK rising edge
t _{HOLD}	5		ns	SDI valid after SCLK rising edge
t _{SDO}		40	ns	SCLK falling edge to SDO/SDIO output transition
t _R ⁴		20	ns	SDO/SDIO output low to output high transition
t _F ⁴		20	ns	SDO/SDIO output high to output low transition

¹ The \overline{CS} , SCLK, SDI, and SDO pins are not internally pulled up or down; they must be driven for proper operation.

² Limits are based on characterization results; not production tested.

³ The timing values are measured corresponding to the input thresholds (V_{IL} and V_{IH}) given in Table 9.

⁴ Output rise and fall times are measured with a capacitive load of 150 pF.

I²C

With \overline{CS} tied high to $V_{DD\ I/O}$, the ADXL346 is in I²C mode, requiring a simple 2-wire connection as shown in Figure 40. The ADXL346 conforms to the *UM10204 I²C-Bus Specification and User Manual*, Rev. 03—19 June 2007, available from NXP Semiconductors. It supports standard (100 kHz) and fast (400 kHz) data transfer modes if the bus parameters given in Table 11 and Table 12 are met. Single- or multiple-byte reads/writes are supported, as shown in Figure 41. With the ALT ADDRESS pin (Pin 7) high, the 7-bit I²C address for the device is 0x1D, followed by the R/W bit. This translates to 0x3A for a write and 0x3B for a read. An alternate I²C address of 0x53 (followed by the R/W bit) can be chosen by grounding the ALT ADDRESS pin. This translates to 0xA6 for a write and 0xA7 for a read.

There are no internal pull-up or pull-down resistors for any unused pins; therefore, there is no known state or default state for the \overline{CS} or ALT ADDRESS pin if left floating or unconnected. It is required that the \overline{CS} pin be connected to $V_{DD\ I/O}$ and that the ALT ADDRESS pin be connected to either $V_{DD\ I/O}$ or GND when using I²C.

Due to communication speed limitations, the maximum output data rate when using 400 kHz I²C is 800 Hz and scales linearly with a change in the I²C communication speed. For example, using I²C at 100 kHz would limit the maximum ODR to 200 Hz. Operation at an output data rate above the recommended maximum may result in an undesirable effect on the acceleration data, including missing samples or additional noise.

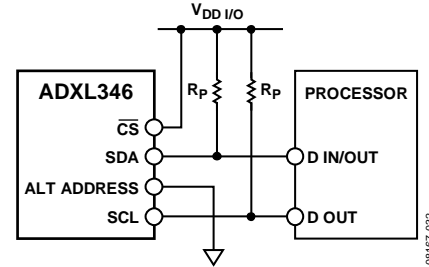


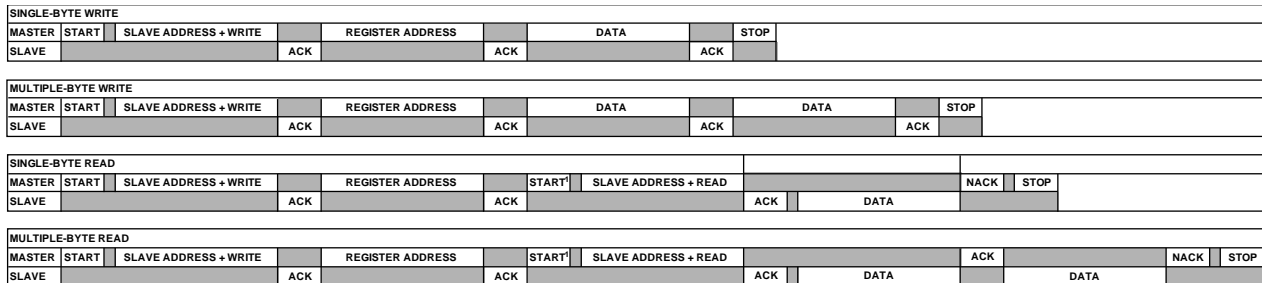
Figure 40. I²C Connection Diagram (Address 0x53)

If other devices are connected to the same I²C bus, the nominal operating voltage level of these other devices cannot exceed $V_{DD\ I/O}$ by more than 0.3 V. External pull-up resistors, R_p , are necessary for proper I²C operation. Refer to the *UM10204 I²C-Bus Specification and User Manual*, Rev. 03—19 June 2007, when selecting pull-up resistor values to ensure proper operation.

Table 11. I²C Digital Input/Output

Parameter	Test Conditions	Limit ¹		Unit
		Min	Max	
Digital Input				
Low Level Input Voltage (V_{IL})			$0.3 \times V_{DD\ I/O}$	V
High Level Input Voltage (V_{IH})		$0.7 \times V_{DD\ I/O}$		V
Low Level Input Current (I_{IL})	$V_{IN} = V_{DD\ I/O}$		0.1	μA
High Level Input Current (I_{IH})	$V_{IN} = 0\ V$	-0.1		μA
Digital Output				
Low Level Output Voltage (V_{OL})	$V_{DD\ I/O} < 2\ V, I_{OL} = 3\ mA$ $V_{DD\ I/O} \geq 2\ V, I_{OL} = 3\ mA$		$0.2 \times V_{DD\ I/O}$	V
Low Level Output Current (I_{OL})	$V_{OL} = V_{OL, max}$	3	400	mV mA
Pin Capacitance	$f_{IN} = 1\ MHz, V_{IN} = 2.6\ V$		8	pF

¹ Limits are based on characterization results; not production tested.



¹ THIS START IS EITHER A RESTART OR A STOP FOLLOWED BY A START.

NOTES

1. THE SHADED AREAS REPRESENT WHEN THE DEVICE IS LISTENING.

Figure 41. I²C Device Addressing

Table 12. I²C Timing (T_A = 25°C, V_S = 2.6 V, V_{DD I/O} = 1.8 V)

Parameter	Limit ^{1,2}		Unit	Description
	Min	Max		
f _{SCL}		400	kHz	SCL clock frequency
t ₁	2.5		μs	SCL cycle time
t ₂	0.6		μs	t _{HIGH} , SCL high time
t ₃	1.3		μs	t _{LOW} , SCL low time
t ₄	0.6		μs	t _{HD, STA} , start/repeated start condition hold time
t ₅	100		ns	t _{SU, DAT} , data setup time
t ₆ ^{3,4,5,6}	0	0.9	μs	t _{HD, DAT} , data hold time
t ₇	0.6		μs	t _{SU, STA} , setup time for repeated start
t ₈	0.6		μs	t _{SU, STO} , stop condition setup time
t ₉	1.3		μs	t _{BUF} , bus-free time between a stop condition and a start condition
t ₁₀		300	ns	t _R , rise time of both SCL and SDA when receiving
t ₁₁	0	300	ns	t _R , rise time of both SCL and SDA when receiving or transmitting
		250	ns	t _F , fall time of SDA when receiving
		400	ns	t _F , fall time of both SCL and SDA when transmitting
C _B		400	pF	Capacitive load for each bus line

¹ Limits are based on characterization results, with f_{SCL} = 400 kHz and a 3 mA sink current; not production tested.

² All values referred to the V_{IH} and the V_{IL} levels given in Table 11.

³ t₆ is the data hold time that is measured from the falling edge of SCL. It applies to data in transmission and acknowledge.

⁴ A transmitting device must internally provide an output hold time of at least 300 ns for the SDA signal (with respect to V_{IH,min} of the SCL signal) to bridge the undefined region of the falling edge of SCL.

⁵ The maximum t₆ value must be met only if the device does not stretch the low period (t₃) of the SCL signal.

⁶ The maximum value for t₆ is a function of the clock low time (t₃), the clock rise time (t₁₀), and the minimum data setup time (t_{5(min)}). This value is calculated as t_{6(max)} = t₃ - t₁₀ - t_{5(min)}.

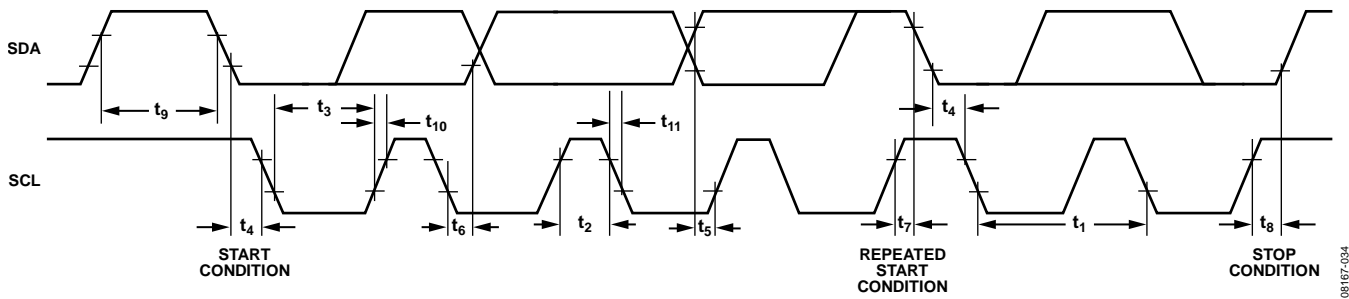


Figure 42. I²C Timing Diagram

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INTERRUPTS

The ADXL346 provides two output pins for driving interrupts: INT1 and INT2. Both interrupt pins are push-pull, low impedance pins with the output specifications listed in Table 13. The default configuration of the interrupt pins is active high. This can be changed to active low by setting the INT_INVERT bit (Bit D5) in the DATA_FORMAT (Address 0x31) register. All functions can be used simultaneously, with the only limiting feature being that some functions may need to share interrupt pins.

Interrupts are enabled by setting the appropriate bit in the INT_ENABLE register (Address 0x2E) and are mapped to either the INT1 or INT2 pin based on the contents of the INT_MAP register (Address 0x2F). When initially configuring the interrupt pins, it is recommended that the functions and interrupt mapping be done before enabling the interrupts. When changing the configuration of an interrupt, it is recommended that the interrupt be disabled first, by clearing the bit corresponding to that function in the INT_ENABLE register, and then the function be reconfigured before enabling the interrupt again. Configuration of the functions while the interrupts are disabled helps to prevent the accidental generation of an interrupt before it is desired.

The interrupt functions are latched and cleared by either reading the DATA_X, DATA_Y, and DATA_Z registers (Address 0x32 to Address 0x37) until the interrupt condition is no longer valid for the data-related interrupts or by reading the INT_SOURCE register (Address 0x30) for the remaining interrupts. This section describes the interrupts that can be set in the INT_ENABLE register and monitored in the INT_SOURCE register.

DATA_READY Bit

The DATA_READY bit is set when new data is available and is cleared when no new data is available.

SINGLE_TAP Bit

The SINGLE_TAP bit is set when a single acceleration event that is greater than the value in the THRESH_TAP register (Address 0x1D) occurs for less time than is specified in the DUR register (Address 0x21).

DOUBLE_TAP Bit

The DOUBLE_TAP bit is set when two acceleration events that are greater than the value in the THRESH_TAP register (Address 0x1D) occur for less time than is specified in the DUR register (Address 0x21). The second tap starts after the time specified by the latent register (Address 0x22) but within the time specified in the window register (Address 0x23). See the Tap Detection section for more details.

Activity Bit

The activity bit is set when acceleration greater than the value stored in the THRESH_ACT register (Address 0x24) is experienced on any participating axis, as set by the ACT_INACT_CTL register (Address 0x27).

Inactivity Bit

The inactivity bit is set when acceleration of less than the value stored in the THRESH_INACT register (Address 0x25) is experienced for more time than is specified in the TIME_INACT register (Address 0x26) on all participating axes, as set by the ACT_INACT_CTL register (Address 0x27). The maximum value for TIME_INACT is 255 sec.

FREE_FALL Bit

The FREE_FALL bit is set when acceleration of less than the value stored in the THRESH_FF register (Address 0x28) is experienced for more time than is specified in the TIME_FF register (Address 0x29) on all axes (logical AND). The FREE_FALL interrupt differs from the inactivity interrupt as follows: all axes always participate and are logically AND'ed, the timer period is much smaller (1.28 sec maximum), and the mode of operation is always dc-coupled.

Watermark Bit

The watermark bit is set when the number of samples in FIFO equals the value stored in the samples bits (Register FIFO_CTL, Address 0x38). The watermark bit is cleared automatically when FIFO is read, and the content returns to a value below the value stored in the samples bits.

Table 13. Interrupt Pin Digital Output

Parameter	Test Conditions	Limit ¹		Unit
		Min	Max	
Digital Output				
Low Level Output Voltage (V_{OL})	$I_{OL} = 300 \mu A$		$0.2 \times V_{DD I/O}$	V
High Level Output Voltage (V_{OH})	$I_{OH} = -150 \mu A$	$0.8 \times V_{DD I/O}$		V
Low Level Output Current (I_{OL})	$V_{OL} = V_{OL, max}$	300		μA
High Level Output Current (I_{OH})	$V_{OH} = V_{OH, min}$		-150	μA
Pin Capacitance	$f_{IN} = 1 \text{ MHz}, V_{IN} = 2.6 \text{ V}$		8	pF
Rise/Fall Time				
Rise Time (t_R) ²	$C_{LOAD} = 150 \text{ pF}$		210	ns
Fall Time (t_F) ³	$C_{LOAD} = 150 \text{ pF}$		150	ns

¹ Limits are based on characterization results; not production tested.

² Rise time is measured as the transition time from $V_{OL, max}$ to $V_{OH, min}$ of the interrupt pin.

³ Fall time is measured as the transition time from $V_{OH, min}$ to $V_{OL, max}$ of the interrupt pin.

Overflow Bit

The overflow bit is set when new data replaces unread data. The precise operation of the overflow function depends on the FIFO mode. In bypass mode, the overflow bit is set when new data replaces unread data in the DATA_X, DATA_Y, and DATA_Z registers (Address 0x32 to Address 0x37). In all other modes, the overflow bit is set when FIFO is filled. The overflow bit is automatically cleared when the contents of FIFO are read.

Orientation Bit

The orientation bit is set when the orientation of the accelerometer changes from a valid orientation to a different valid orientation. An interrupt is not generated, however, if the orientation of the accelerometer changes from a valid orientation to an invalid orientation, or from a valid orientation to an invalid orientation and then back to the same valid orientation. An invalid orientation is defined as an orientation within the dead zone, or the region of hysteresis. This region helps to prevent rapid orientation change due to noise when the accelerometer orientation is close to the boundary between two valid orientations.

The orientations that are valid for the interrupt depend on which mode, 2D or 3D, is linked to the orientation interrupt. The mode is selected with the INT_3D bit (Bit D3) in the ORIENT_CONF register (Address 0x3B). See the Register 0x3B—ORIENT_CONF (Read/Write) section for more details on how to enable the orientation interrupt.

FIFO

The ADXL346 contains technology for an embedded memory management system with 32-level FIFO that can be used to minimize host processor burden. This buffer has four modes: bypass, FIFO, stream, and trigger (see Table 22). Each mode is selected by the settings of the FIFO_MODE bits (Bits[D7:D6]) in the FIFO_CTL register (Address 0x38).

Bypass Mode

In bypass mode, FIFO is not operational and, therefore, remains empty.

FIFO Mode

In FIFO mode, data from measurements of the x-, y-, and z-axes are stored in FIFO. When the number of samples in FIFO equals the level specified in the samples bits of the FIFO_CTL register (Address 0x38), the watermark interrupt is set. FIFO continues accumulating samples until it is full (32 samples from measurements of the x-, y-, and z-axes) and then stops collecting data. After FIFO stops collecting data, the device continues to operate; therefore, features such as tap detection can be used after FIFO is full. The watermark interrupt continues to occur until the number of samples in FIFO is less than the value stored in the samples bits of the FIFO_CTL register.

Stream Mode

In stream mode, data from measurements of the x-, y-, and z-axes are stored in FIFO. When the number of samples in FIFO equals the level specified in the samples bits of the FIFO_CTL register (Address 0x38), the watermark interrupt is set. FIFO continues accumulating samples and holds the latest 32 samples from measurements of the x-, y-, and z-axes, discarding older data as new data arrives. The watermark interrupt continues occurring until the number of samples in FIFO is less than the value stored in the samples bits of the FIFO_CTL register.

Trigger Mode

In trigger mode, FIFO accumulates samples, holding the latest 32 samples from measurements of the x-, y-, and z-axes. After a trigger event occurs and an interrupt is sent to the INT1 or INT2 pin (determined by the trigger bit in the FIFO_CTL register), FIFO keeps the last n samples (where n is the value specified by the samples bits in the FIFO_CTL register) and then operates in FIFO mode, collecting new samples only when FIFO is not full. A delay of at least 5 μ s should be present between the trigger event occurring and the start of reading data from the FIFO to allow the FIFO to discard and retain the necessary samples. Additional trigger events cannot be recognized until the trigger mode is reset. To reset the trigger mode, set the device to bypass mode and then set the device back to trigger mode. Note that the FIFO data should be read first because placing the device into bypass mode clears FIFO.

Retrieving Data from FIFO

The FIFO data is read through the DATA_X, DATA_Y, and DATA_Z registers (Address 0x32 to Address 0x37). When the FIFO is in FIFO, stream, or trigger mode, reads to the DATA_X, DATA_Y, and DATA_Z registers read data stored in the FIFO. Each time data is read from the FIFO, the oldest x-, y-, and z-axes data are placed into the DATA_X, DATA_Y, and DATA_Z registers.

If a single-byte read operation is performed, the remaining bytes of data for the current FIFO sample are lost. Therefore, all axes of interest should be read in a burst (or multiple-byte) read operation. To ensure that the FIFO has completely popped (that is, that new data has completely moved into the DATA_X, DATA_Y, and DATA_Z registers), there must be at least 5 μ s between the end of reading the data registers and the start of a new read of the FIFO or a read of the FIFO_STATUS register (Address 0x39). The end of reading a data register is signified by the transition of data from Register 0x37 to Register 0x38 or by the \overline{CS} pin going high.

For SPI operation at 1.6 MHz or less, the register addressing portion of the transmission is a sufficient delay to ensure that the FIFO has completely popped. For SPI operation greater than 1.6 MHz, it is necessary to deassert the \overline{CS} pin to ensure a total delay of 5 μ s; otherwise, the delay will not be sufficient. The total delay necessary for 5 MHz operation is at most 3.4 μ s. This is not a concern when using I²C mode because the communication rate is low enough to ensure a sufficient delay between FIFO reads.

SELF-TEST

The ADXL346 incorporates a self-test feature that effectively tests its mechanical and electronic systems simultaneously. When the self-test function is enabled (via the SELF_TEST bit (Bit D7 in the DATA_FORMAT register, Address 0x31), an electrostatic force is exerted on the mechanical sensor. This electrostatic force moves the mechanical sensing element in the same manner as acceleration would, and it is additive to the acceleration experienced by the device. This added electrostatic force results in an output change in the x-, y-, and z-axes. Because the electrostatic force is proportional to V_s^2 , the output change varies with V_s . This effect is shown in Figure 43.

The scale factors listed in Table 14 can be used to adjust the expected self-test output limits for different supply voltages, V_s . The self-test feature of the ADXL346 also exhibits a bimodal behavior. However, the limits listed in Table 1 and Table 15 to Table 18 are valid for both potential self-test values due to bimodality. Use of the self-test feature at data rates less than 100 Hz or at 1600 Hz may yield values outside these limits. Therefore, the part must be in normal power operation (LOW_POWER bit = 0 in the BW_RATE register, Address 0x2C) and be placed into a data rate of 100 Hz through 800 Hz or 3200 Hz for the self-test function to operate correctly.

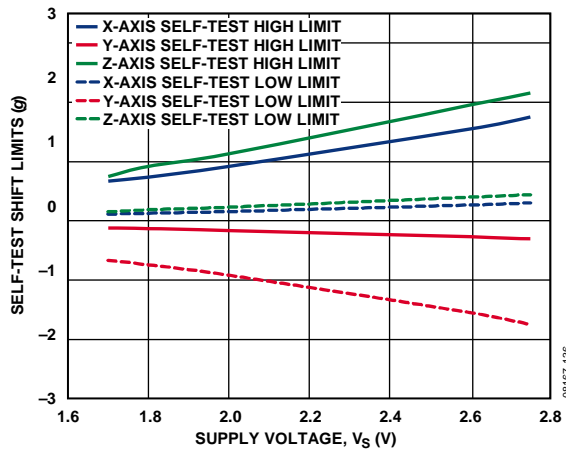


Figure 43. Self-Test Output Change Limits vs. Supply Voltage

Table 14. Self-Test Output Scale Factors for Different Supply Voltages, V_s

Supply Voltage, V_s	X-, Y-Axes	Z-Axis
1.70 V	0.43	0.38
1.80 V	0.48	0.47
2.00 V	0.59	0.58
2.60 V	1.00	1.00
2.75 V	1.13	1.11

Table 15. Self-Test Output in LSB for ± 2 g, 10-Bit or Full Resolution ($T_A = 25^\circ\text{C}$, $V_s = 2.6$ V, $V_{DD I/O} = 1.8$ V)

Axis	Min	Max	Unit
X	70	400	LSB
Y	-400	-70	LSB
Z	100	500	LSB

Table 16. Self-Test Output in LSB for ± 4 g, 10-Bit Resolution ($T_A = 25^\circ\text{C}$, $V_s = 2.6$ V, $V_{DD I/O} = 1.8$ V)

Axis	Min	Max	Unit
X	35	200	LSB
Y	-200	-35	LSB
Z	50	250	LSB

Table 17. Self-Test Output in LSB for ± 8 g, 10-Bit Resolution ($T_A = 25^\circ\text{C}$, $V_s = 2.6$ V, $V_{DD I/O} = 1.8$ V)

Axis	Min	Max	Unit
X	17	100	LSB
Y	-100	-17	LSB
Z	25	125	LSB

Table 18. Self-Test Output in LSB for ± 16 g, 10-Bit Resolution ($T_A = 25^\circ\text{C}$, $V_s = 2.6$ V, $V_{DD I/O} = 1.8$ V)

Axis	Min	Max	Unit
X	8	50	LSB
Y	-50	-8	LSB
Z	12	63	LSB

REGISTER MAP

Table 19. Register Map

Address		Name	Type	Reset Value	Description
Hex	Dec				
0x00	0	DEVID	R	11100110	Device ID.
0x01 to 0x1C	1 to 28	Reserved			Reserved. Do not access.
0x1D	29	THRESH_TAP	R/W	00000000	Tap threshold.
0x1E	30	OFSX	R/W	00000000	X-axis offset.
0x1F	31	OFSY	R/W	00000000	Y-axis offset.
0x20	32	OFSZ	R/W	00000000	Z-axis offset.
0x21	33	DUR	R/W	00000000	Tap duration.
0x22	34	Latent	R/W	00000000	Tap latency.
0x23	35	Window	R/W	00000000	Tap window.
0x24	36	THRESH_ACT	R/W	00000000	Activity threshold.
0x25	37	THRESH_INACT	R/W	00000000	Inactivity threshold.
0x26	38	TIME_INACT	R/W	00000000	Inactivity time.
0x27	39	ACT_INACT_CTL	R/W	00000000	Axis enable control for activity and inactivity detection.
0x28	40	THRESH_FF	R/W	00000000	Free-fall threshold.
0x29	41	TIME_FF	R/W	00000000	Free-fall time.
0x2A	42	TAP_AXES	R/W	00000000	Axis control for single tap/double tap.
0x2B	43	ACT_TAP_STATUS	R	00000000	Source of single tap/double tap.
0x2C	44	BW_RATE	R/W	00001010	Data rate and power mode control.
0x2D	45	POWER_CTL	R/W	00000000	Power-saving features control.
0x2E	46	INT_ENABLE	R/W	00000000	Interrupt enable control.
0x2F	47	INT_MAP	R/W	00000000	Interrupt mapping control.
0x30	48	INT_SOURCE	R	00000010	Source of interrupts.
0x31	49	DATA_FORMAT	R/W	00000000	Data format control.
0x32	50	DATA0	R	00000000	X-Axis Data 0.
0x33	51	DATA1	R	00000000	X-Axis Data 1.
0x34	52	DATAY0	R	00000000	Y-Axis Data 0.
0x35	53	DATAY1	R	00000000	Y-Axis Data 1.
0x36	54	DATAZ0	R	00000000	Z-Axis Data 0.
0x37	55	DATAZ1	R	00000000	Z-Axis Data 1.
0x38	56	FIFO_CTL	R/W	00000000	FIFO control.
0x39	57	FIFO_STATUS	R	00000000	FIFO status.
0x3A	58	TAP_SIGN	R	00000000	Sign and source for single tap/double tap.
0x3B	59	ORIENT_CONF	R/W	00100101	Orientation configuration.
0x3C	60	Orient	R	00000000	Orientation status.

REGISTER DEFINITIONS**Register 0x00—DEVID (Read Only)**

D7	D6	D5	D4	D3	D2	D1	D0
1	1	1	0	0	1	1	0

The DEVID register holds a fixed device ID code of 0xE6 (346 octal).

Register 0x1D—THRESH_TAP (Read/Write)

The THRESH_TAP register is eight bits and holds the threshold value for tap interrupts. The data format is unsigned, so the magnitude of the tap event is compared with the value in THRESH_TAP for normal tap detection. For information on improved tap detection, refer to the Improved Tap Detection section. The scale factor is 62.5 mg/LSB (that is, 0xFF = +16 g). A value of 0 may result in undesirable behavior if single-tap/double-tap interrupts are enabled.

Register 0x1E, Register 0x1F, Register 0x20—OFSX, OFSY, OFSZ (Read/Write)

The OFSX, OFSY, and OFSZ registers are each eight bits and offer user-set offset adjustments in twos complement format with a scale factor of 15.6 mg/LSB (that is, 0x7F = 2 g). The values stored in the offset registers are automatically added to the acceleration data, and the resulting value is stored in the output data registers. For additional information regarding offset calibration and the use of the offset registers, refer to the Offset Calibration section.

Register 0x21—DUR (Read/Write)

The DUR register is eight bits and contains an unsigned time value representing the maximum time that an event must be above the THRESH_TAP threshold to qualify as a tap event. For information on improved tap detection, refer to the Improved Tap Detection section. The scale factor is 625 μ s/LSB. A value of 0 disables the single-tap/double-tap functions.

Register 0x22—Latent (Read/Write)

The latent register is eight bits and contains an unsigned time value representing the wait time from the detection of a tap event to the start of the time window (defined by the window register) during which a possible second tap event can be detected. For information on improved tap detection, refer to the Improved Tap Detection section. The scale factor is 1.25 ms/LSB. A value of 0 disables the double-tap function.

Register 0x23—Window (Read/Write)

The window register is eight bits and contains an unsigned time value representing the amount of time after the expiration of the latency time (determined by the latent register) during which a second valid tap can begin. For information on improved tap detection, refer to the Improved Tap Detection section. The scale factor is 1.25 ms/LSB. A value of 0 disables the double-tap function.

Register 0x24—THRESH_ACT (Read/Write)

The THRESH_ACT register is eight bits and holds the threshold value for detecting activity. The data format is unsigned, so the magnitude of the activity event is compared with the value in the THRESH_ACT register. The scale factor is 62.5 mg/LSB. A value of 0 may result in undesirable behavior if the activity interrupt is enabled.

Register 0x25—THRESH_INACT (Read/Write)

The THRESH_INACT register is eight bits and holds the threshold value for detecting inactivity. The data format is unsigned, so the magnitude of the inactivity event is compared with the value in the THRESH_INACT register. The scale factor is 62.5 mg/LSB. A value of 0 may result in undesirable behavior if the inactivity interrupt is enabled.

Register 0x26—TIME_INACT (Read/Write)

The TIME_INACT register is eight bits and contains an unsigned time value representing the amount of time that acceleration must be less than the value in the THRESH_INACT register for inactivity to be declared. The scale factor is 1 sec/LSB. Unlike the other interrupt functions, which use unfiltered data (see the Threshold section), the inactivity function uses filtered output data. At least one output sample must be generated for the inactivity interrupt to be triggered. This results in the function appearing unresponsive if the TIME_INACT register is set to a value less than the time constant of the output data rate. A value of 0 results in an interrupt when the output data is less than the value in the THRESH_INACT register.

Register 0x27—ACT_INACT_CTL (Read/Write)

D7	D6	D5	D4
ACT ac/dc	ACT_X enable	ACT_Y enable	ACT_Z enable
D3	D2	D1	D0
INACT ac/dc	INACT_X enable	INACT_Y enable	INACT_Z enable

ACT AC/DC and INACT AC/DC Bits

A setting of 0 selects dc-coupled operation, and a setting of 1 enables ac-coupled operation. In dc-coupled operation, the current acceleration magnitude is compared directly with THRESH_ACT and THRESH_INACT to determine whether activity or inactivity is detected.

In ac-coupled operation for activity detection, the acceleration value at the start of activity detection is taken as a reference value. New samples of acceleration are then compared to this reference value, and if the magnitude of the difference exceeds the THRESH_ACT value, the device triggers an activity interrupt.

Similarly, in ac-coupled operation for inactivity detection, a reference value is used for comparison and is updated whenever the device exceeds the inactivity threshold. After the reference value is selected, the device compares the magnitude of the difference between the reference value and the current acceleration with THRESH_INACT. If the difference is less than the value in THRESH_INACT for the time in TIME_INACT, the device is considered inactive and the inactivity interrupt is triggered.

ACT_x Enable Bits and INACT_x Enable Bits

A setting of 1 enables x-, y-, or z-axis participation in detecting activity or inactivity. A setting of 0 excludes the selected axis from participation. If all axes are excluded, the function is disabled. For activity detection, all participating axes are logically ORed, causing the activity function to trigger when any of the participating axes exceeds the threshold. For inactivity detection, all participating axes are logically ANDed, causing the inactivity function to trigger only if all participating axes are below the threshold for the specified period of time.

Register 0x28—THRESH_FF (Read/Write)

The THRESH_FF register is eight bits and holds the threshold value, in unsigned format, for free-fall detection. The acceleration on all axes is compared with the value in THRESH_FF to determine if a free-fall event occurred. The scale factor is 62.5 mg/LSB. Note that a value of 0 mg may result in undesirable behavior if the free-fall interrupt is enabled. Values between 300 mg and 600 mg (0x05 to 0x09) are recommended.

Register 0x29—TIME_FF (Read/Write)

The TIME_FF register is eight bits and stores an unsigned time value representing the minimum time that the value of all axes must be less than THRESH_FF to generate a free-fall interrupt. The scale factor is 5 ms/LSB. A value of 0 may result in undesirable behavior if the free-fall interrupt is enabled. Values between 100 ms and 350 ms (0x14 to 0x46) are recommended.

Register 0x2A—TAP_AXES (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	Improved tap	Suppress	TAP_X enable	TAP_Y enable	TAP_Z enable

Improved Tap Bit

The improved tap bit is used to enable improved tap detection. This mode of operation improves tap detection by performing an ac-coupled differential comparison of the output acceleration data. The improved tap detection is performed on the same output data available in the DATA_X, DATA_Y, and DATA_Z registers. Due to the dependency on the output data rate and the ac-coupled differential measurement, the threshold and timing values for single taps and double taps must be adjusted for improved tap detection. For further explanation of improved tap detection, see the Improved Tap Detection section. Improved tap is enabled by setting the improved tap bit to a value of 1 and is disabled by clearing the bit to a value of 0.

Suppress Bit

Setting the suppress bit suppresses double-tap detection if acceleration greater than the value in THRESH_TAP is present between taps. See the Tap Detection section for more details.

TAP_x Enable Bits

A setting of 1 in the TAP_X enable, TAP_Y enable, or TAP_Z enable bit enables x-, y-, or z-axis participation in tap detection. A setting of 0 excludes the selected axis from participation in tap detection.

Register 0x2B—ACT_TAP_STATUS (Read Only)

D7	D6	D5	D4	D3	D2	D1	D0
0	ACT_X source	ACT_Y source	ACT_Z source	Asleep	TAP_X source	TAP_Y source	TAP_Z source

ACT_x Source and TAP_x Source Bits

These bits indicate the first axis involved in a tap or activity event. A setting of 1 corresponds to involvement in the event, and a setting of 0 corresponds to no involvement. When new data is available, these bits are not cleared but are overwritten by the new data. The ACT_TAP_STATUS register should be read before clearing the interrupt. Disabling an axis from participation clears the corresponding source bit when the next activity or single-tap/double-tap event occurs.

Asleep Bit

A setting of 1 in the asleep bit indicates that the part is asleep, and a setting of 0 indicates that the part is not asleep. This bit toggles only if the device is configured for autosleep. See the Register 0x2D—POWER_CTL (Read/Write) section for more information on autosleep mode.

Register 0x2C—BW_RATE (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	LOW_POWER	Rate			

LOW_POWER Bit

A setting of 0 in the LOW_POWER bit selects normal operation, and a setting of 1 selects reduced power operation, which is associated with somewhat higher noise (see the Power Modes section for details).

Rate Bits

These bits select the device bandwidth and output data rate (see Table 7 and Table 8 for details). The default value is 0x0A, which translates to a 100 Hz output data rate. An output data rate should be selected that is appropriate for the communication protocol and frequency selected. Selecting too high of an output data rate with a low communication speed results in samples being discarded.

Register 0x2D—POWER_CTL (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
0	0	Link	AUTO_SLEEP	Measure	Sleep	Wakeup	

Link Bit

A setting of 1 in the link bit with both the activity and inactivity functions enabled delays the start of the activity function until inactivity is detected. After activity is detected, inactivity detection begins, preventing the detection of activity. This bit serially links the activity and inactivity functions. When this bit is set to 0, the inactivity and activity functions are concurrent. Additional information can be found in the Link Mode section.

When clearing the link bit, it is recommended that the part be placed into standby mode and then set back to measurement mode with a subsequent write. This is done to ensure that the device is properly biased if sleep mode is manually disabled; otherwise, the first few samples of data after the link bit is cleared may have additional noise, especially if the device was asleep when the bit was cleared.

AUTO_SLEEP Bit

If the link bit is set, a setting of 1 in the AUTO_SLEEP bit enables the autosleep functionality. In this mode, the ADXL346 automatically switches to sleep mode if the inactivity function is enabled and inactivity is detected (that is, when acceleration is below the THRESH_INACT value for at least the time indicated by TIME_INACT). If activity is also enabled, the ADXL346 automatically wakes up from sleep after detecting activity and returns to operation at the output data rate set in the BW_RATE register. A setting of 0 in the AUTO_SLEEP bit disables automatic switching to sleep mode. See the description of the sleep bit in this section for more information on sleep mode.

If the link bit is not set, the AUTO_SLEEP feature is disabled, and setting the AUTO_SLEEP bit does not have any impact on device operation. Refer to the Link Bit section or the Link Mode section for more information about using the link feature.

When clearing the AUTO_SLEEP bit, it is recommended that the part be placed into standby mode and then set back to measurement mode with a subsequent write. This is done to ensure that the device is properly biased if sleep mode is manually disabled; otherwise, the first few samples of data after the AUTO_SLEEP bit is cleared may have additional noise, especially if the device was asleep when the bit was cleared.

Measure Bit

A setting of 0 in the measure bit places the part into standby mode, and a setting of 1 places the part into measurement mode. The ADXL346 powers up in standby mode with minimum power consumption.

Sleep Bit

A setting of 0 in the sleep bit puts the part into the normal mode of operation, and a setting of 1 places the part into sleep mode. Sleep mode suppresses DATA_READY, stops transmission of data to FIFO, and switches the sampling rate to one specified by the wakeup bits. In sleep mode, only the activity function can be used. While the DATA_READY interrupt is suppressed, the output data registers are still updated at the sampling rate set by the wakeup bits.

When clearing the sleep bit, it is recommended that the part be placed into standby mode and then set back to measurement mode with a subsequent write. This is done to ensure that the device is properly biased if sleep mode is manually disabled; otherwise, the first few samples of data after the sleep bit is cleared may have additional noise, especially if the device was asleep when the bit was cleared.

Wakeup Bits

These bits control the frequency of readings in sleep mode as described in Table 20.

Table 20. Frequency of Readings in Sleep Mode

Setting		Frequency (Hz)
D1	D0	
0	0	8
0	1	4
1	0	2
1	1	1

Register 0x2E—INT_ENABLE (Read/Write)

D7 DATA_READY	D6 SINGLE_TAP	D5 DOUBLE_TAP	D4 Activity
D3 Inactivity	D2 FREE_FALL	D1 Watermark	D0 Overrun/ orientation

Setting bits in this register to a value of 1 enables their respective functions to generate interrupts, whereas a value of 0 prevents the functions from generating interrupts. The DATA_READY, watermark, and overrun/orientation bits enable only the interrupt output; the functions are always enabled. It is recommended that interrupts be configured before enabling their outputs.

Register 0x2F—INT_MAP (Read/Write)

D7 DATA_READY	D6 SINGLE_TAP	D5 DOUBLE_TAP	D4 Activity
D3 Inactivity	D2 FREE_FALL	D1 Watermark	D0 Overrun/ orientation

Bits set to 0 in this register send their respective interrupts to the INT1 pin, whereas bits set to 1 send their respective interrupts to the INT2 pin. All selected interrupts for a given pin are ORed.

Register 0x30—INT_SOURCE (Read Only)

D7 DATA_READY	D6 SINGLE_TAP	D5 DOUBLE_TAP	D4 Activity
D3 Inactivity	D2 FREE_FALL	D1 Watermark	D0 Overrun/ orientation

Bits set to 1 in this register indicate that their respective functions have triggered an event, whereas bits set to 0 indicate that the corresponding events have not occurred. The DATA_READY, watermark, and overrun/orientation bits are always set if the corresponding events occur, regardless of the INT_ENABLE register settings, and are cleared by reading data from the DATA_X, DATA_Y, and DATA_Z registers. The DATA_READY and watermark bits may require multiple reads, as indicated in the FIFO mode descriptions in the FIFO section. Other bits, and the corresponding interrupts, including orientation if enabled, are cleared by reading the INT_SOURCE register.

Register 0x31—DATA_FORMAT (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
SELF_TEST	SPI	INT_INVERT	0	FULL_RES	Justify	Range	

The DATA_FORMAT register controls the presentation of data to Register 0x32 through Register 0x37. All data, except that for the $\pm 16 g$ range, must be clipped to avoid rollover.

SELF_TEST Bit

A setting of 1 in the SELF_TEST bit applies a self-test force to the sensor, causing a shift in the output data. A value of 0 disables the self-test force.

SPI Bit

A value of 1 in the SPI bit sets the device to 3-wire SPI mode, and a value of 0 sets the device to 4-wire SPI mode.

INT_INVERT Bit

A value of 0 in the INT_INVERT bit sets the interrupts to active high, and a value of 1 sets the interrupts to active low.

FULL_RES Bit

When this bit is set to a value of 1, the device is in full resolution mode, where the output resolution increases with the g range set by the range bits to maintain a 4 mg/LSB scale factor. When the FULL_RES bit is set to 0, the device is in 10-bit mode, and the range bits determine the maximum g range and scale factor.

Justify Bit

A setting of 1 in the justify bit selects left-justified (MSB) mode, and a setting of 0 selects right-justified mode with sign extension.

Range Bits

These bits set the g range as described in Table 21.

Table 21. g Range Setting

Setting		g Range
D1	D0	
0	0	$\pm 2 g$
0	1	$\pm 4 g$
1	0	$\pm 8 g$
1	1	$\pm 16 g$

Register 0x32 to Register 0x37—DATA_{X0}, DATA_{X1}, DATA_{Y0}, DATA_{Y1}, DATA_{Z0}, DATA_{Z1} (Read Only)

These six bytes (Register 0x32 to Register 0x37) are eight bits each and hold the output data for each axis. Register 0x32 and Register 0x33 hold the output data for the x-axis, Register 0x34 and Register 0x35 hold the output data for the y-axis, and Register 0x36 and Register 0x37 hold the output data for the z-axis. The output data is twos complement, with DATA_{X0} as the least significant byte

and DATA_{X1} as the most significant byte, where x represents X, Y, or Z. The DATA_FORMAT register (Address 0x31) controls the format of the data. It is recommended that a multiple-byte read of all registers be performed to prevent a change in data between reads of sequential registers.

Register 0x38—FIFO_CTL (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
FIFO_MODE		Trigger	Samples				

FIFO_MODE Bits

These bits set the FIFO mode, as described in Table 22.

Table 22. FIFO Modes

Setting		Mode	Function
D7	D6		
0	0	Bypass	FIFO is bypassed.
0	1	FIFO	FIFO collects up to 32 values and then stops collecting data, collecting new data only when FIFO is not full.
1	0	Stream	FIFO holds the last 32 data values. When FIFO is full, the oldest data is overwritten with newer data.
1	1	Trigger	When triggered by the trigger bit, FIFO holds the last data samples before the trigger event and then continues to collect data until FIFO is full. New data is collected only when FIFO is not full.

Trigger Bit

A value of 0 in the trigger bit links the trigger event of trigger mode to INT1, and a value of 1 links the trigger event to INT2.

Samples Bits

The function of these bits depends on the FIFO mode selected (see Table 23). Entering a value of 0 in the samples bits immediately sets the watermark bit in the INT_SOURCE register (Address 0x30), regardless of which FIFO mode is selected. Undesirable operation may occur if a value of 0 is used for the samples bits when trigger mode is used.

Table 23. Samples Bits Functions

FIFO Mode	Samples Bits Function
Bypass	None.
FIFO	Specifies how many FIFO entries are needed to trigger a watermark interrupt.
Stream	Specifies how many FIFO entries are needed to trigger a watermark interrupt.
Trigger	Specifies how many FIFO samples are retained in the FIFO buffer before a trigger event.

Register 0x39—FIFO_STATUS (Read Only)

D7	D6	D5	D4	D3	D2	D1	D0
FIFO_TRIG	0	Entries					

FIFO_TRIG Bit

A 1 in the FIFO_TRIG bit corresponds to a trigger event occurring, and a 0 means that a FIFO trigger event has not occurred.

Entries Bits

These bits report how many data values are stored in FIFO. Access to collect the data from FIFO is provided through the DATA_X, DATA_Y, and DATA_Z registers. FIFO reads must be done in burst or multiple-byte mode because each FIFO level is cleared after any read (single- or multiple-byte) of FIFO. FIFO stores a maximum of 32 entries, which equates to a maximum of 33 entries available at any given time because an additional entry is available at the output filter of the device.

Register 0x3A—TAP_SIGN (Read Only)

D7	D6	D5	D4	D3	D2	D1	D0
0	XSIGN	YSIGN	ZSIGN	0	XTAP	YTAP	ZTAP

xSIGN Bits

These bits indicate the sign of the first axis involved in a tap event. A setting of 1 corresponds to acceleration in the negative direction, and a setting of 0 corresponds to acceleration in the positive direction. These bits update only when a new single-tap/double-tap event is detected, and only the axes enabled in the TAP_AXES register (Address 0x2A) are updated. The TAP_SIGN register should be read before clearing the interrupt. See the Tap Sign section for more details.

xTAP Bits

These bits indicate the first axis involved in a tap event. A setting of 1 corresponds to involvement in the event, and a setting of 0 corresponds to no involvement. When new data is available, these bits are not cleared but are overwritten by the new data. The TAP_SIGN register should be read before clearing the interrupt. Disabling an axis from participation clears the corresponding source bit when the next single-tap/double-tap event occurs.

Register 0x3B—ORIENT_CONF (Read/Write)

D7	D6	D5	D4	D3	D2	D1	D0
INT_ORIENT	Dead zone			INT_3D	Divisor		

INT_ORIENT Bit

Setting the INT_ORIENT bit enables the orientation interrupt. A value of 1 overrides the overrun function of the device and replaces overrun in the INT_MAP (Address 0x2F), INT_ENABLE (Address 0x2E), and INT_SOURCE (Address 0x30) registers with the orientation function. After setting the INT_ORIENT bit, the orientation bits in the INT_MAP and INT_ENABLE registers must be configured to map the orientation interrupt to INT1 or INT2 and to enable generation of the interrupt to the pin.

An orientation interrupt is generated whenever the orientation status for the mode selected by the INT_3D bit changes in the orient register (Address 0x3C). The orientation interrupt is cleared by reading the INT_SOURCE register. Clearing the INT_ORIENT bit, or the orientation bit in the INT_ENABLE register (Address 0x2E), disables and clears the interrupt.

Writing to the BW_RATE register (Address 0x2C) or placing the part into standby mode resets the orientation feature, clearing the orientation filter and the interrupt. However, resetting the orientation feature also resets the orientation status in the orient register (Address 0x3C) and, therefore, causes an interrupt to be generated when the next output sample is available if the present orientation is not the default orientation. A value of 0 for the INT_ORIENT bit disables generation of the orientation interrupt and permits the use of the overrun function.

Dead Zone Bits

These bits determine the region between two adjacent orientations, where the orientation is considered invalid and is not updated. A value of 0 may result in undesirable behavior when the orientation is close to the bisector between two adjacent regions. The dead zone angle is determined by these bits, as described in Table 24. See the Orientation Sensing section for more details.

Table 24. Dead Zone and Divisor Codes

Decimal	Binary	Dead Zone Angle (Degrees)	Divisor Bandwidth (Hz)
0	000	5.1	ODR/9
1	001	10.2	ODR/22
2	010	15.2	ODR/50
3	011	20.4	ODR/100
4	100	25.5	ODR/200
5	101	30.8	ODR/400
6	110	36.1	ODR/800
7	111	41.4	ODR/1600

INT_3D Bit

If the orientation interrupt is enabled, the INT_3D bit determines whether 2D or 3D orientation detection generates an interrupt. A value of 0 generates an interrupt only if the 2D orientation changes from a valid 2D orientation to a different valid 2D orientation. A value of 1 generates an interrupt only if the 3D orientation changes from a valid 3D orientation to a different valid 3D orientation.

Divisor Bits

These bits set the bandwidth of the filter used to low-pass filter the measured acceleration for stable orientation sensing. The divisor bandwidth is determined by these bits, as detailed in Table 24, where ODR is the output data rate set in the BW_RATE register (Address 0x2C). See the Orientation Sensing section for more details.

Register 0x3C—Orient (Read Only)

D7	D6	D5	D4	D3	D2	D1	D0
0	V2	2D_ORIENT		V3	3D_ORIENT		

Vx Bits

These bits show the validity of the 2D (V2) and 3D (V3) orientations. A value of 1 corresponds to the orientation being valid. A value of 0 means that the orientation is invalid because the current orientation is in the dead zone.

xD_ORIENT Bits

These bits represent the current 2D (2D_ORIENT) and 3D (3D_ORIENT) orientations of the accelerometer. If the orientation interrupt is enabled, this register is read to determine the orientation of the device when the interrupt occurs. Because this register updates with each new sample of acceleration data, it should be read at the time of the orientation interrupt to ensure that the orientation change that caused the interrupt has been identified. Orientation values are shown in Table 25 and Table 26. See the Orientation Sensing section for more details.

Writing to the BW_RATE register (Address 0x2C) or placing the part into standby mode resets the orientation feature, clearing the orientation filter and the orientation status. An orientation interrupt (if enabled) results from these actions if the orientation during the next output sample is different from the default value (+X for 2D orientation detection and undefined for 3D orientation).

Table 25. 2D Orientation Codes

Decimal	Binary	Orientation	Dominant Axis
0	00	Portrait positive	+X
1	01	Portrait negative	-X
2	10	Landscape positive	+Y
3	11	Landscape negative	-Y

Table 26. 3D Orientation Codes

Decimal	Binary	Orientation	Dominant Axis
3	011	Front	+X
4	100	Back	-X
2	010	Left	+Y
5	101	Right	-Y
1	001	Top	+Z
6	110	Bottom	-Z

APPLICATIONS INFORMATION

POWER SUPPLY DECOUPLING

A 1 μF tantalum capacitor (C_S) at V_S and a 0.1 μF ceramic capacitor ($C_{I/O}$) at $V_{DD\ I/O}$ placed close to the ADXL346 supply pins is recommended to adequately decouple the accelerometer from noise on the power supply. If additional decoupling is necessary, a resistor or ferrite bead, no larger than 100 Ω , in series with V_S may be helpful. Additionally, increasing the bypass capacitance on V_S to a 10 μF tantalum capacitor in parallel with a 0.1 μF ceramic capacitor may also improve noise.

Care should be taken to ensure that the connection from the ADXL346 ground to the power supply ground has low impedance because noise transmitted through ground has an effect similar to noise transmitted through V_S . It is recommended that V_S and $V_{DD\ I/O}$ be separate supplies to minimize digital clock noise on the V_S supply. If this is not possible, additional filtering of the supplies as previously mentioned may be necessary.

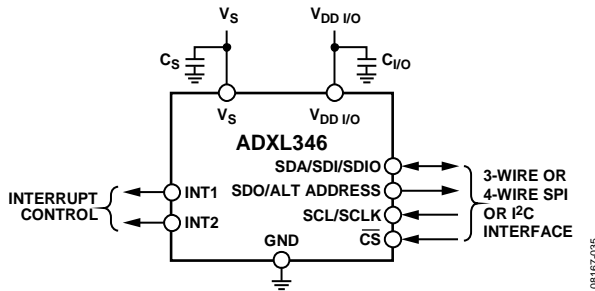


Figure 44. Applications Diagram

MECHANICAL CONSIDERATIONS FOR MOUNTING

The ADXL346 should be mounted on the PCB in a location close to a hard mounting point of the PCB to the case. Mounting the ADXL346 at an unsupported PCB location, as shown in Figure 45, may result in large, apparent measurement errors due to undamped PCB vibration. Locating the accelerometer near a hard mounting point ensures that any PCB vibration at the accelerometer is above the accelerometer’s mechanical sensor resonant frequency and, therefore, effectively invisible to the accelerometer. Multiple mounting points, close to the sensor, and/or a thicker PCB also help to reduce the effect of system resonance on the performance of the sensor.

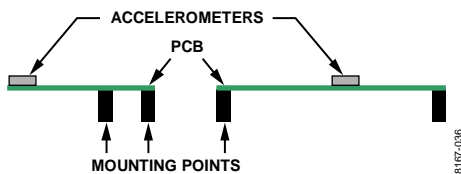


Figure 45. Incorrectly Placed Accelerometers

TAP DETECTION

The tap interrupt function is capable of detecting either single or double taps. The following parameters are shown in Figure 46 for a valid single-tap event and a valid double-tap event:

- The tap detection threshold is defined by the THRESH_TAP register (Address 0x1D).
- The maximum tap duration time is defined by the DUR register (Address 0x21).
- The tap latency time is defined by the latent register (Address 0x22) and is the waiting period from the end of the first tap until the start of the time window when a second tap can be detected, which is determined by the value in the window register (Address 0x23).
- The interval after the latency time (set by the latent register) is defined by the window register. Although a second tap must begin after the latency time has expired, it need not finish before the end of the time defined by the window register.

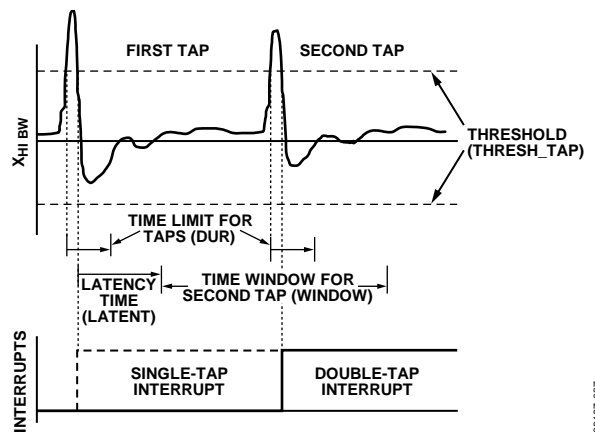


Figure 46. Tap Interrupt Function with Valid Single and Double Taps

If only the single-tap function is in use, the single-tap interrupt is triggered when the acceleration goes below the threshold, as long as DUR has not been exceeded. If both single and double-tap functions are in use, the single-tap interrupt is triggered when the double-tap event has been either validated or invalidated.

Several events can occur to invalidate the second tap of a double-tap event. First, if the suppress bit in the TAP_AXES register (Address 0x2A) is set, any acceleration spike above the threshold during the latency time (set by the latent register) invalidates the double-tap detection, as shown in Figure 47.

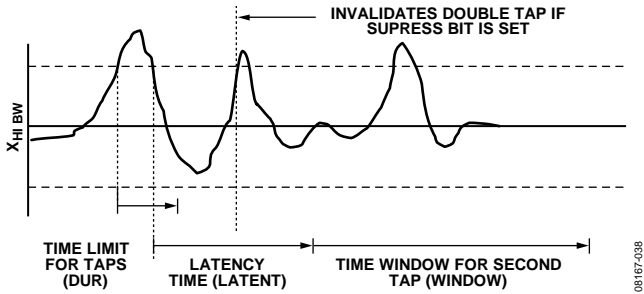


Figure 47. Double-Tap Event Invalid Due to High g Event When the Suppress Bit Is Set

A double-tap event can also be invalidated if acceleration above the threshold is detected at the start of the time window for the second tap (set by the window register (Address 0x23)). This results in an invalid double tap at the start of this window, as shown in Figure 48. Additionally, a double-tap event can be invalidated if an acceleration exceeds the time limit for taps (set by the DUR register (Address 0x21)), resulting in an invalid double tap at the end of the DUR time limit for the second tap event, also shown in Figure 48.

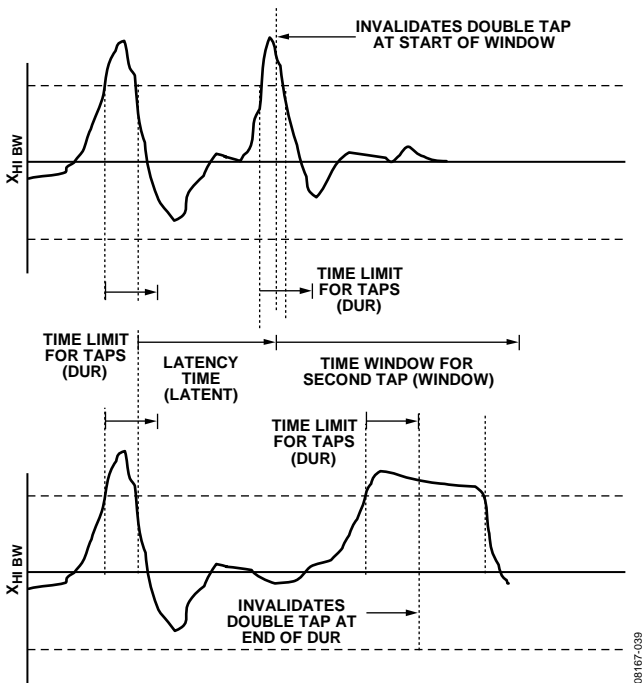


Figure 48. Tap Interrupt Function with Invalid Double Taps

Single taps, double taps, or both can be detected by setting the respective bits in the INT_ENABLE register (Address 0x2E). Control over participation of each of the three axes in single-tap/double-tap detection is exerted by setting the appropriate bits in the TAP_AXES register (Address 0x2A). For the double-tap function to operate, both the latent and window registers must be set to a nonzero value.

Every mechanical system has somewhat different single-tap/double-tap responses based on the mechanical characteristics of the system. Therefore, some experimentation with values for the

DUR, latent, window, and THRESH_TAP registers is required. In general, a good starting point is to set the DUR register to a value greater than 0x10 (10 ms), the latent register to a value greater than 0x10 (20 ms), the window register to a value greater than 0x40 (80 ms), and the THRESH_TAP register to a value greater than 0x30 (3 g). Setting a very low value in the latent, window, or THRESH_TAP register may result in an unpredictable response due to the accelerometer picking up echoes of the tap inputs.

After a tap interrupt has been received, the first axis to exceed the THRESH_TAP level is reported in the ACT_TAP_STATUS register (Address 0x2B). This register is never cleared but is overwritten with new data.

IMPROVED TAP DETECTION

Improved tap detection is enabled by setting the improved tap bit of the TAP_AXES register (Address 0x2A). When improved tap detection is enabled, the filtered output data corresponding to the output data rate set in the BW_RATE register (Address 0x2C) is processed to determine if a tap event occurred. In addition, an ac-coupled differential measurement is used. This results in the timing values and threshold values for improved tap detection being different from those used for normal tap detection.

When improved tap detection is used, new values must be determined based on test results. In general, no timing values (in the DUR, latent, or window registers) should be set that are less than the time step resolution set by the output data rate. The threshold value for improved tap detection can typically be set much lower than the threshold for normal tap detection. The value used depends on the value in the BW_RATE register and should be determined through system testing. Refer to the Threshold section for more details.

TAP SIGN

A negative sign is produced by experiencing a negative acceleration, which corresponds to tapping on the positive face of the device for the desired axis. The positive face of the device is the face such that movement in that direction is positive acceleration. For example, tapping on the face that corresponds to the +X direction, labeled as front in Figure 49, results in a negative sign for the x-axis. Tapping on the face labeled as left in Figure 49 results in a negative sign for the y-axis, and tapping on the face labeled top results in a negative sign for the z-axis. Conversely, tapping on the back, right, or bottom side results in positive signs for the corresponding axes.

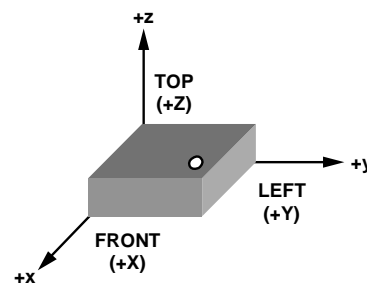


Figure 49. 3D Orientation with Coordinate System

THRESHOLD

The lower output data rates are achieved by decimating a common sampling frequency inside the device. The activity, free-fall, and single-tap/double-tap detection functions without improved tap enabled are performed using undecimated data. As the bandwidth of the output data varies with the data rate and is lower than the bandwidth of the undecimated data, the high frequency and high *g* data that is used to determine activity, free-fall, and single-tap/double-tap events may not be present if the output of the accelerometer is examined. This may result in functions triggering when acceleration data does not appear to meet the conditions set by the user for the corresponding function.

LINK MODE

The function of the link bit is to reduce the number of activity interrupts that the processor must service by setting the device to look for activity only after inactivity. For proper operation of this feature, the processor must still respond to the activity and inactivity interrupts by reading the INT_SOURCE register (Address 0x30) and, therefore, clearing the interrupts. If an activity interrupt is not cleared, the part cannot go into autosleep mode. The asleep bit in the ACT_TAP_STATUS register (Address 0x2B) indicates whether the part is asleep.

SLEEP MODE VS. LOW POWER MODE

In applications where a low data rate and low power consumption are desired (at the expense of noise performance), it is recommended that low power mode be used. The use of low power mode preserves the functionality of the DATA_READY interrupt and the FIFO for postprocessing of the acceleration data. Sleep mode, while offering a low data rate and power consumption, is not intended for data acquisition.

However, when sleep mode is used in conjunction with the AUTO_SLEEP mode and the link mode, the part can automatically switch to a low power, low sampling rate mode when inactivity is detected. To prevent the generation of redundant inactivity interrupts, the inactivity interrupt is automatically disabled and activity is enabled. When the ADXL346 is in sleep mode, the host processor can also be placed into sleep mode or low power mode to save significant system power. Once activity is detected, the accelerometer automatically switches back to the original data rate of the application and provides an activity interrupt that can be used to wake up the host processor. Similar to when inactivity occurs, detection of activity events is disabled and inactivity is enabled.

OFFSET CALIBRATION

Accelerometers are mechanical structures containing elements that are free to move. These moving parts can be very sensitive to mechanical stresses, much more so than solid-state electronics. The 0 *g* bias or offset is an important accelerometer metric because it defines the baseline for measuring acceleration. Additional stresses can be applied during assembly of a system containing an accelerometer. These stresses can come from, but are not limited to, component soldering, board stress during mounting,

and application of any compounds on or over the component. If calibration is deemed necessary, it is recommended that calibration be performed after system assembly to compensate for these effects.

A simple method of calibration is to measure the offset while assuming that the sensitivity of the ADXL346 is as specified in Table 1. The offset can then be automatically accounted for by using the built-in offset registers (Register 0x1E, Register 0x1F, and Register 0x20). This results in the data acquired from the DATA_X, DATA_Y, and DATA_Z registers (Address 0x32 to Address 0x37) already compensating for any offset.

In a no-turn or single-point calibration scheme, the part is oriented such that one axis, typically the *z*-axis, is in the 1 *g* field of gravity and the remaining axes, typically the *x*- and *y*-axes, are in a 0 *g* field. The output is then measured by taking the average of a series of samples. The number of samples averaged is a choice of the system designer, but a recommended starting point is 0.1 sec worth of data for data rates of 100 Hz or greater. This corresponds to 10 samples at the 100 Hz data rate. For data rates of less than 100 Hz, it is recommended that at least 10 samples be averaged together. These values are stored as X_{0g} , Y_{0g} , and Z_{+1g} for the 0 *g* measurements on the *x*- and *y*-axes and the 1 *g* measurement on the *z*-axis, respectively.

The values measured for X_{0g} and Y_{0g} correspond to the offset of the *x*- and *y*-axes, and compensation is done by subtracting those values from the output of the accelerometer to obtain the actual acceleration:

$$X_{ACTUAL} = X_{MEAS} - X_{0g}$$

$$Y_{ACTUAL} = Y_{MEAS} - Y_{0g}$$

Because the *z*-axis measurement is done in a 1 *g* field, a no-turn or single-point calibration scheme assumes an ideal sensitivity, S_z , for the *z*-axis. This is subtracted from Z_{+1g} to attain the *z*-axis offset, which is then subtracted from future measured values to obtain the actual value:

$$Z_{0g} = Z_{1g} - S_z$$

$$Z_{ACTUAL} = Z_{MEAS} - Z_{0g}$$

The ADXL346 can automatically compensate the output for offset by using the offset registers (Register 0x1E, Register 0x1F, and Register 0x20). These registers contain an 8-bit, twos complement value that is automatically added to all measured acceleration values, and the result is then placed into the DATA_X, DATA_Y, and DATA_Z registers. Because the value placed in an offset register is additive, a negative value is placed into the register to eliminate a positive offset and vice versa for a negative offset. The register has a scale factor of 15.6 mg/LSB and is independent of the selected *g* range.

As an example, assume that the ADXL346 is placed into full-resolution mode with a sensitivity of typically 256 LSB/*g*. The part is oriented such that the *z*-axis is in the field of gravity and the outputs of the *x*-, *y*-, and *z*-axes are measured as +10 LSB, -13 LSB, and +9 LSB, respectively. Using the previous equations, X_{0g} is +10 LSB, Y_{0g} is -13 LSB, and Z_{0g} is +9 LSB. Each LSB of

output in full-resolution is 3.9 mg or one-quarter of an LSB of the offset register.

Because the offset register is additive, the 0 g values are negated and rounded to the nearest LSB of the offset register:

$$X_{\text{OFFSET}} = -\text{Round}(10/4) = -3 \text{ LSB}$$

$$Y_{\text{OFFSET}} = -\text{Round}(-13/4) = 3 \text{ LSB}$$

$$Z_{\text{OFFSET}} = -\text{Round}(9/4) = -2 \text{ LSB}$$

These values are programmed into the OFSX, OFSY, and OFXZ registers, respectively, as 0xFD, 0x03, and 0xFE. As with all registers in the ADXL346, the offset registers do not retain the value written into them when power is removed from the part. Power-cycling the ADXL346 returns the offset registers to their default value of 0x00.

Because the no-turn or single-point calibration method assumes an ideal sensitivity in the z-axis, any error in the sensitivity results in offset error. For instance, if the actual sensitivity was 250 LSB/g in the previous example, the offset would be 15 LSB, not 9 LSB. To help minimize this error, an additional measurement point can be used with the z-axis in a 0 g field, and the 0 g measurement can be used in the Z_{ACTUAL} equation.

USING SELF-TEST

The self-test change is defined as the difference between the acceleration output of an axis with self-test enabled and the acceleration output of the same axis with self-test disabled (see Endnote 8 of Table 1). This definition assumes that the sensor does not move between these two measurements, because if the sensor moves, a non-self-test related shift corrupts the test.

Proper configuration of the ADXL346 is also necessary for an accurate self-test measurement. The part should be set with a data rate of 100 Hz through 800 Hz, or 3200 Hz. This is done by ensuring that a value of 0x0A through 0x0D, or 0x0F is written into the rate bits (Bit D3 through Bit D0) in the BW_RATE register (Address 0x2C). The part also must be placed into normal power operation by ensuring that the LOW_POWER bit (Bit D4) in the BW_RATE register is cleared (LOW_POWER bit = 0) for accurate self-test measurements. It is recommended that the part be set to full-resolution, 16 g mode to ensure that there is sufficient dynamic range for the entire self-test shift. This is done by setting the FULL_RES bit (Bit D3) and writing a value of 0x03 to the range bits (Bit D1 and Bit D0) of the DATA_FORMAT register (Address 0x31). This results in a high dynamic range for measurement and a 3.9 mg/LSB scale factor.

After the part is configured for accurate self-test measurement, several samples of acceleration data for the x-, y-, and z-axes should be retrieved from the sensor and averaged together. The number of samples averaged is a choice of the system designer, but a recommended starting point is 0.1 sec worth of data for data rates of 100 Hz or greater. This corresponds to 10 samples at the 100 Hz data rate. For data rates of less than 100 Hz, it is

recommended that at least 10 samples be averaged together. The averaged values should be stored and labeled appropriately as the self-test disabled data, that is, $X_{\text{ST_OFF}}$, $Y_{\text{ST_OFF}}$, and $Z_{\text{ST_OFF}}$.

Next, self-test should be enabled by setting Bit D7 of the DATA_FORMAT register (Address 0x31). The output needs some time (about four samples) to settle once self-test is enabled. After allowing the output to settle, several samples of acceleration data for the x-, y-, and z-axes should be taken again and averaged. It is recommended that the same number of samples be taken for this average as was previously taken. These averaged values should again be stored and labeled appropriately as the value with self-test enabled, that is, $X_{\text{ST_ON}}$, $Y_{\text{ST_ON}}$, and $Z_{\text{ST_ON}}$. Self-test can then be disabled by clearing Bit D7 of the DATA_FORMAT register (Address 0x31).

With the stored values for self-test enabled and disabled, the self-test change is as follows:

$$X_{\text{ST}} = X_{\text{ST_ON}} - X_{\text{ST_OFF}}$$

$$Y_{\text{ST}} = Y_{\text{ST_ON}} - Y_{\text{ST_OFF}}$$

$$Z_{\text{ST}} = Z_{\text{ST_ON}} - Z_{\text{ST_OFF}}$$

Because the measured output for each axis is expressed in LSBs, X_{ST} , Y_{ST} , and Z_{ST} are also expressed in LSBs. These values can be converted to g's of acceleration by multiplying each value by the 3.9 mg/LSB scale factor, if configured for full-resolution mode. Additionally, Table 15 through Table 18 correspond to the self-test range converted to LSBs and can be compared with the measured self-test change when operating at a V_S of 2.6 V. For other voltages, the minimum and maximum self-test output values should be adjusted based on (multiplied by) the scale factors shown in Table 14. If the part was placed into ± 2 g, 10-bit or full-resolution mode, the values listed in Table 15 should be used. Although the fixed 10-bit mode or a range other than 16 g can be used, a different set of values, as indicated in Table 16 through Table 18, would need to be used. Using a range below 8 g may result in insufficient dynamic range and should be considered when selecting the range of operation for measuring self-test.

If the self-test change is within the valid range, the test is considered successful. Generally, a part is considered to pass if the minimum magnitude of change is achieved. However, a part that changes by more than the maximum magnitude is not necessarily a failure.

Another effective method for using the self-test to verify accelerometer functionality is to toggle the self-test at a certain rate and then perform an FFT on the output. The FFT should have a corresponding tone at the frequency the self-test was toggled. This methodology removes the dependency of the test on supply voltage and on self-test magnitude, which can vary within a rather wide range.

ORIENTATION SENSING

The orientation function of the ADXL346 reports both 2D and 3D orientation concurrently through the orient register (Address 0x3C). The V2 and V3 bits (Bit D6 and Bit D3 in the orient register) report the validity of the 2D and 3D orientation

codes. If V2 or V3 are set, their respective code is a valid orientation. If V2 or V3 are cleared, the orientation of the accelerometer is unknown, such as when the orientation is within the dead zone between valid regions.

For 2D orientation sensing, the relation of the x- and y-axes to gravity is used to determine the accelerometer orientation (see Figure 50 and Table 25). Portrait positive corresponds to the x-axis being most closely aligned to the gravity vector and directed upwards, opposite the gravity vector. Portrait negative is the opposite of portrait positive, with the x-axis pointing downwards along the gravity vector. Landscape positive corresponds to the y-axis being most closely aligned with the gravity vector and directed upwards, away from the gravity vector. Landscape negative is the orientation opposite landscape positive. The dead zone regions are shown in the orientations for portrait positive (+X) and portrait negative (-X) of Figure 50. These regions also exist for landscape positive (+Y) and landscape negative (-Y), as shown in Figure 50.

In 3D orientation, the z-axis is also included. If the accelerometer is placed in a Cartesian coordinate system, as shown in Figure 49 of the Tap Sign section, the top of the device corresponds to the positive z-axis direction, the front of the device corresponds to the positive x-axis direction, and the right side of the device corresponds to the positive y-axis direction.

The states shown in Table 26 correspond to which side of the accelerometer is directed upwards, opposite the gravity vector. As shown in Figure 49, the accelerometer is oriented in the top state. If the device is flipped over such that the top of the device is facing down, toward gravity, the orientation is reported as the bottom state. If the device is adjusted such that the positive x-axis or positive y-axis direction is pointing upwards, away from the gravity vector, the accelerometer reports the orientation as front or left, respectively.

The algorithm to detect orientation change is performed after filtering the output acceleration data to eliminate the effects of high frequency motion. This is performed by using a low-pass filter with a bandwidth set by the divisor bits (ORIENT_CONF register, Address 0x3B). The orientation filter uses the same output data available in the output data registers (Address 0x32 to Address 0x37); therefore, the orient register (Address 0x3C) is updated at the same rate as the data rate that is set in the BW_RATE register (Address 0x2C). Because the output data is used, the bandwidth of the orientation filter depends on the value set in the BW_RATE register, and the divisor bandwidth values in Table 24 are referenced to the selected output data rate.

To eliminate most human motion, such as walking or shaking, the value in the divisor bits (Bits[D2:D0]) of the ORIENT_CONF register (Address 0x3B) should be selected to effectively limit the orientation bandwidth to 1 Hz or 2 Hz. For example, with an output data rate of 100 Hz, a divisor selection of 3 (ODR/100) results in a 1 Hz bandwidth for orientation detection. For best results, it is recommended that an output data rate of ≥ 25 Hz in normal power mode and ≥ 200 Hz in low power operation be used.

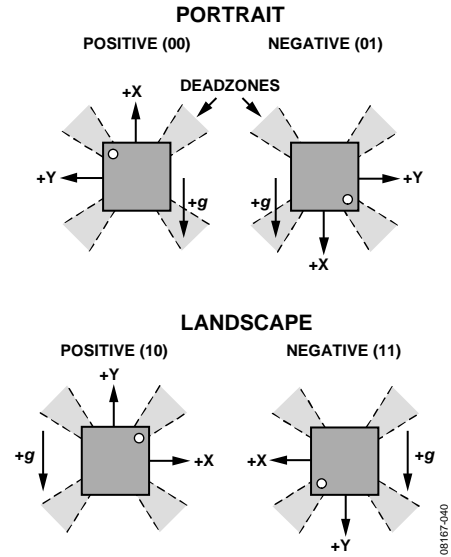


Figure 50. 2D Orientation with Corresponding Codes

The width of the dead zone region between two orientation positions is determined by setting the value of the dead zone bits (Bits[D6:D4]) in the ORIENT_CONF register (Address 0x3B). The dead zone region size can be specified as per the values shown in Table 24. The dead zone angle represents the total angle where the orientation is considered invalid. Therefore, a dead zone of 15.4° corresponds to 7.7° in either direction away from the bisector of two bordering regions. An example with a dead zone region of 15.4° is shown in Figure 51. It should be noted that the values shown in Table 24 correspond to the typical dead zone angle when the gravity vector is completely contained in only two axes (xy, xz, or yz) and should be used only as a starting point. If the device is oriented such that the projection of gravity onto all three axes is nonzero, the effective sensitivity is reduced, causing an increase in the dead zone angle. Therefore, evaluation needs to be performed for specific application uses to determine the optimal setting for the dead zone.

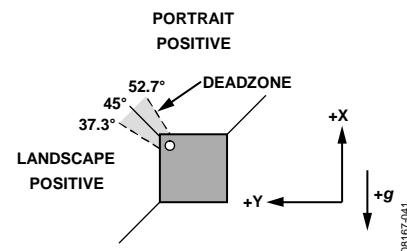


Figure 51. Orientation Showing a 15.4° Dead Zone Region

By setting the INT_ORIENT bit (Bit D7) of the ORIENT_CONF register (Address 0x3B), an interrupt can be generated when the device is placed into a new valid orientation. Only one mode of orientation detection, 2D or 3D, can generate an interrupt at a time. The orientation detection mode is selected by setting or clearing the INT_3D bit (Bit D3) of the ORIENT_CONF register (Address 0x3B). For more details, refer to the Register 0x3B—ORIENT_CONF (Read/Write) section.

Writing to the BW_RATE register or placing the part into standby mode resets the orientation feature, clearing the orientation filter and the orientation status. These actions cause an orientation interrupt (if enabled), however, if the orientation during the next output sample is different from the default value (+X for 2D orientation detection and undefined for 3D orientation).

DATA FORMATTING OF UPPER DATA RATES

Formatting of output data at the 3200 Hz and 1600 Hz output data rates changes depending on the mode of operation (full-resolution or fixed 10-bit) and the selected output range.

When using the 3200 Hz or 1600 Hz output data rates in full-resolution or ±2 g, 10-bit operation, the LSB of the output data-word is always 0. When data is right justified, this corresponds to Bit D0 of the DATAx0 register, as shown in Figure 52. When data is left justified and the part is operating in ±2 g, 10-bit mode,

the LSB of the output data-word is Bit D6 of the DATAx0 register. In full-resolution operation when data is left justified, the location of the LSB changes according to the selected output range. For a range of ±2 g, the LSB is Bit D6 of the DATAx0 register; for ±4 g, Bit D5 of the DATAx0 register; for ±8 g, Bit D4 of the DATAx0 register; and for ±16 g, Bit D3 of the DATAx0 register. This is shown in Figure 53.

The use of 3200 Hz and 1600 Hz output data rates for fixed 10-bit operation in the ±4 g, ±8 g, and ±16 g output ranges provides an LSB that is valid and that changes according to the applied acceleration. Therefore, in these modes of operation, Bit D0 is not always 0 when output data is right justified, and Bit D6 is not always 0 when output data is left justified. Operation at any data rate of 800 Hz or lower also provides a valid LSB in all ranges and modes that change according to the applied acceleration.

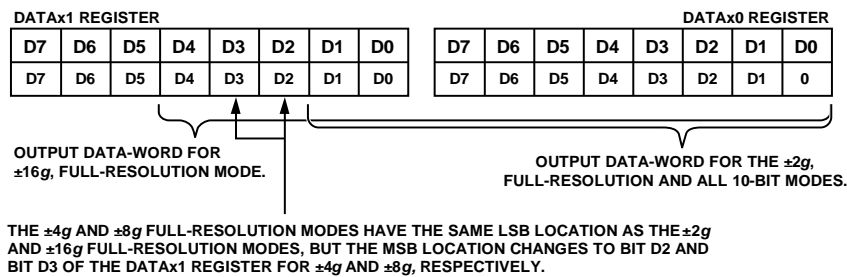


Figure 52. Data Formatting When Output Data Is Right Justified

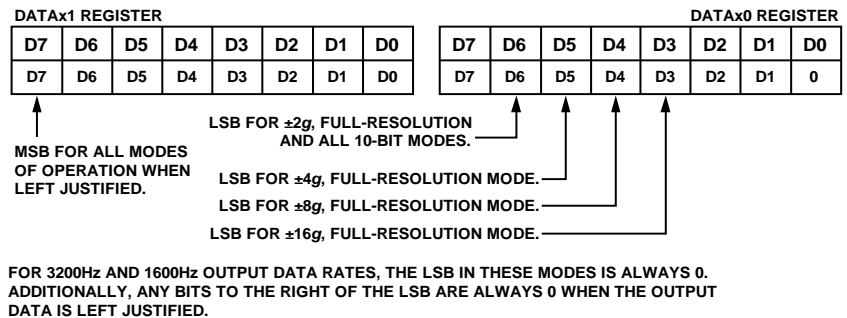


Figure 53. Data Formatting When Output Data Is Left Justified

NOISE PERFORMANCE

The specification of noise shown in Table 1 corresponds to the typical noise performance of the ADXL346 in normal power operation with an output data rate of 100 Hz (LOW_POWER bit = 0, rate = 0x0A in the BW_RATE register, Address 0x2C). For normal power operation at data rates below 100 Hz, the noise of the ADXL346 is equivalent to the noise at 100 Hz ODR in LSBs. For data rates greater than 100 Hz, the noise increases approximately by a factor of $\sqrt{2}$ per doubling of the data rate. For example, at 400 Hz ODR, the noise on the x- and y-axes is typically less than 2 LSB rms, and the noise on the z-axis is typically less than 3 LSB rms.

For low power operation (LOW_POWER bit = 1 in the BW_RATE register, Address 0x2C), the noise of the ADXL346 is constant for all valid data rates shown in Table 8. This value is typically less than 2.83 LSB rms for the x- and y-axes and typically less than 4.25 LSB rms for the z-axis.

The trend of noise performance for both normal power and low power modes of operation of the ADXL346 is shown in Figure 54.

Figure 55 shows the typical Allan deviation for the ADXL346. The 1/f corner of the device, as shown in this figure, is very low, allowing absolute resolution of approximately 100 μg (assuming that there is sufficient integration time). The figure also shows that the noise density is 420 $\mu\text{g}/\sqrt{\text{Hz}}$ for the x- and y-axes and 530 $\mu\text{g}/\sqrt{\text{Hz}}$ for the z-axis.

Figure 56 shows the typical noise performance trend of the ADXL346 over supply voltage. The performance is normalized to the tested and specified supply voltage, $V_S = 2.6 \text{ V}$. The x-axis offers the best noise performance over supply voltage, increasing by typically less than 25% from nominal at a supply voltage of 1.8 V. The performance of the y- and z-axes is comparable, with both axes increasing by typically less than 35% when operating with a supply voltage of 1.8 V. It should be noted, as shown in Figure 54, that the noise on the z-axis is typically higher than that on the y-axis; therefore, although the noise on the z- and y-axes change roughly the same in percentage over supply voltage, the magnitude of change on the z-axis is greater than the magnitude of change on the y-axis.

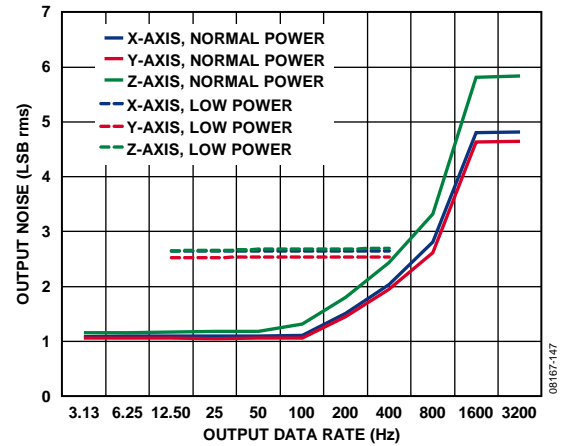


Figure 54. Noise vs. Output Data Rate for Normal and Low Power Modes, Full Resolution (256 LSB/g)

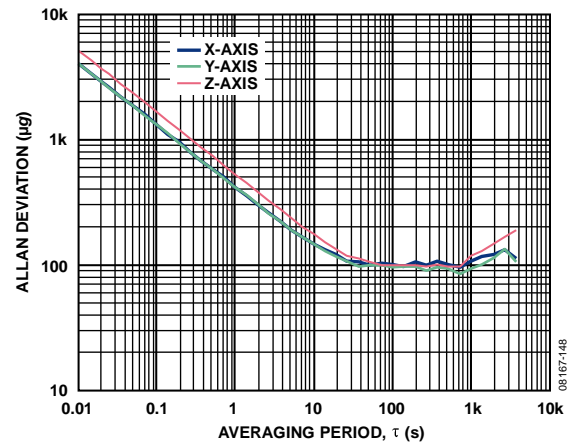


Figure 55. Allan Deviation

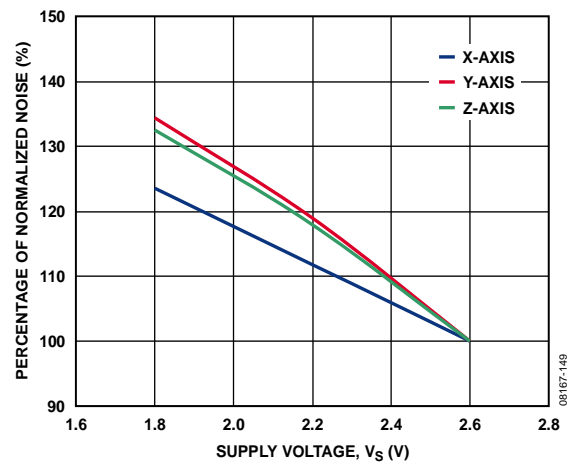


Figure 56. Normalized Noise vs. Supply Voltage

OPERATION AT VOLTAGES OTHER THAN 2.6 V

The ADXL346 is tested and specified at a supply voltage of $V_S = 2.6\text{ V}$; however, it can be powered with a V_S as high as 2.75 V or as low as 1.7 V. Some performance parameters change as the supply voltage changes, including the offset, sensitivity, noise, self-test, and supply current.

Due to minuscule changes in the electrostatic forces as supply voltage is varied, the offset and sensitivity change slightly. When operating at a supply voltage of $V_S = 1.8\text{ V}$, the offset of the x- and y-axes is typically 25 mg higher than at $V_S = 2.6\text{ V}$ operation. The z-axis is typically 20 mg lower when operating at a supply voltage of 1.8 V than when operating at $V_S = 2.6\text{ V}$. Sensitivity on the x- and y-axes typically shifts from a nominal 256 LSB/g (full-resolution or $\pm 2\text{ g}$, 10-bit operation) at $V_S = 2.6\text{ V}$ operation to 250 LSB/g when operating with a supply voltage of 1.8 V. The z-axis sensitivity is unaffected by a change in supply voltage and is the same at $V_S = 1.8\text{ V}$ operation as it is at $V_S = 2.6\text{ V}$ operation. Simple linear interpolation can be used to determine typical shifts in offset and sensitivity at other supply voltages.

Changes in noise performance, self-test response, and supply current are discussed elsewhere throughout the data sheet. For more information about noise performance, review the Noise Performance section. The Self-Test section discusses both the operation of self-test over voltage (a square relationship with the supply voltage) and the conversion of the self-test response in g's to LSBs. Finally, Figure 33 shows the impact of supply voltage on typical current consumption at a 100 Hz output data rate, with all other output data rates following the same trend.

OFFSET PERFORMANCE AT LOWEST DATA RATES

The ADXL346 offers several output data rates and bandwidths designed for a wide range of applications. However, at the lowest data rates, described as those data rates below 6.25 Hz, the offset performance over temperature can vary significantly from the remaining data rates. Figure 57, Figure 58, and Figure 59 show the typical offset performance of the ADXL346 over temperature for data rates of 6.25 Hz and lower. All plots are normalized to the offset at 100 Hz output data rate; therefore, a nonzero value corresponds to additional offset shift due to the temperature for that data rate.

When using the lowest data rates, it is recommended that the operating temperature range of the device be limited to provide minimal offset shift across the operating temperature range. Due to variability between parts, it is also recommended that calibration over temperature be performed if any data rates below 6.25 Hz are in use.

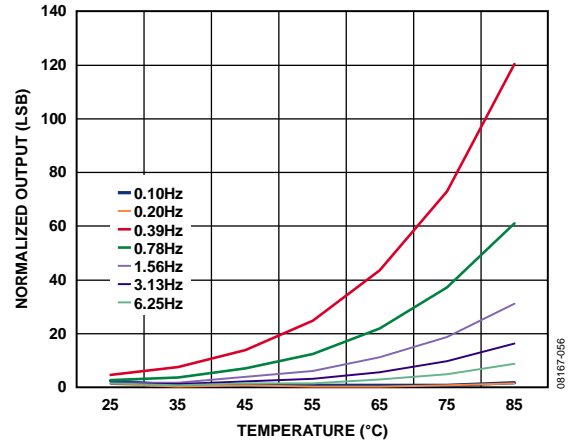


Figure 57. Typical X-Axis Output vs. Temperature at Lower Data Rates, Normalized to 100 Hz Output Data Rate, $V_S = 2.6\text{ V}$

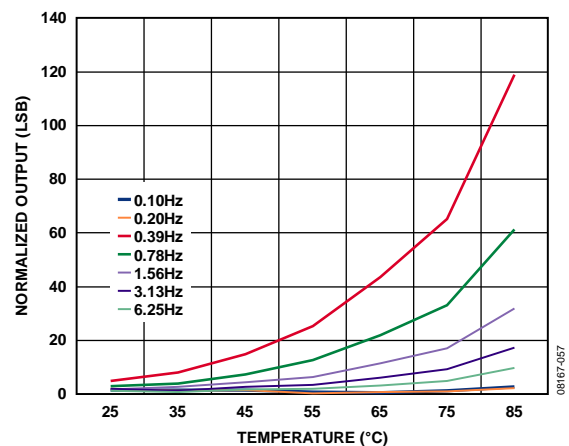


Figure 58. Typical Y-Axis Output vs. Temperature at Lower Data Rates, Normalized to 100 Hz Output Data Rate, $V_S = 2.6\text{ V}$

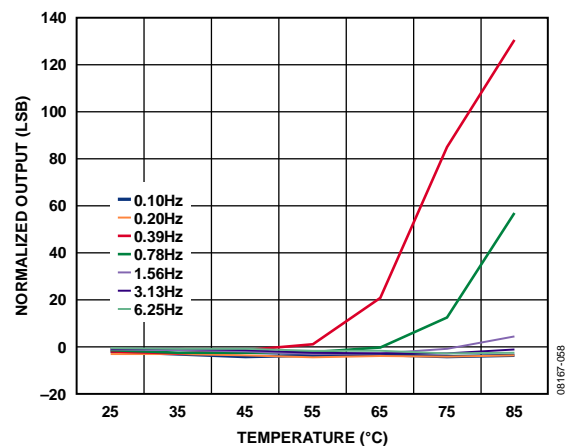


Figure 59. Typical Z-Axis Output vs. Temperature at Lower Data Rates, Normalized to 100 Hz Output Data Rate, $V_S = 2.6\text{ V}$

AXES OF ACCELERATION SENSITIVITY

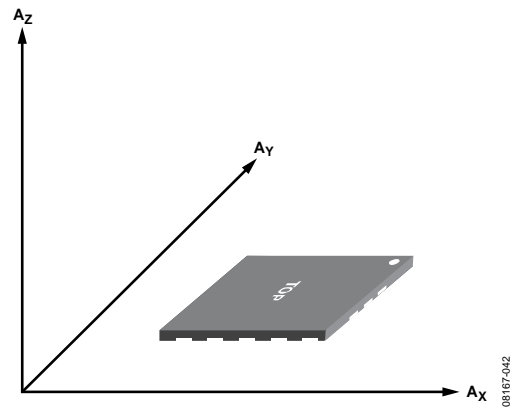


Figure 60. Axes of Acceleration Sensitivity (Corresponding Output Increases When Accelerated Along the Sensitive Axis)

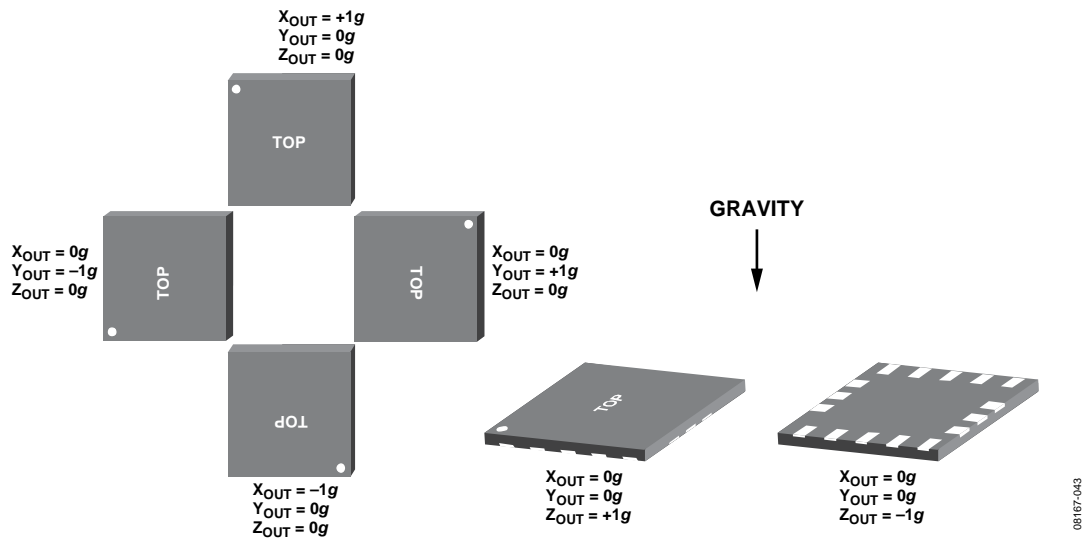


Figure 61. Output Response vs. Orientation to Gravity

LAYOUT AND DESIGN RECOMMENDATIONS

Figure 62 shows the recommended printed wiring board land pattern. Figure 63 and Table 27 provide details about the recommended soldering profile.

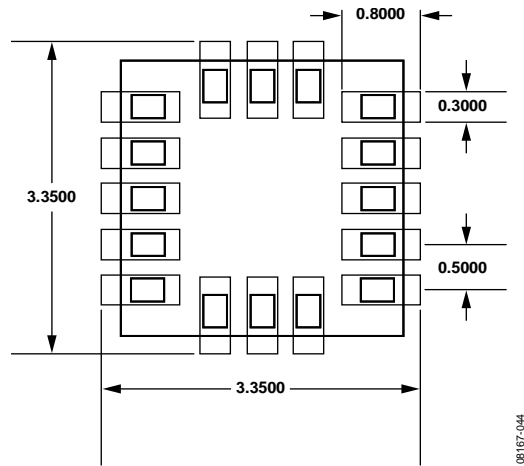


Figure 62. Recommended Printed Wiring Board Land Pattern (Dimensions shown in millimeters)

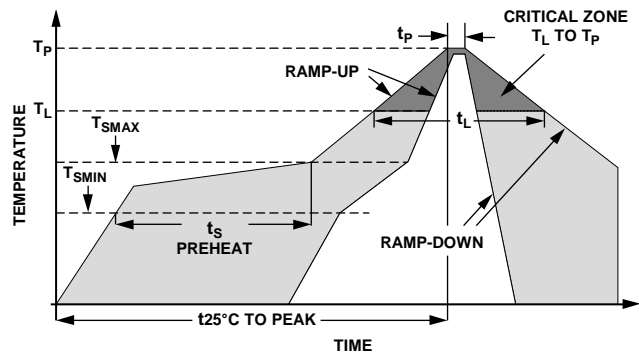


Figure 63. Recommended Soldering Profile

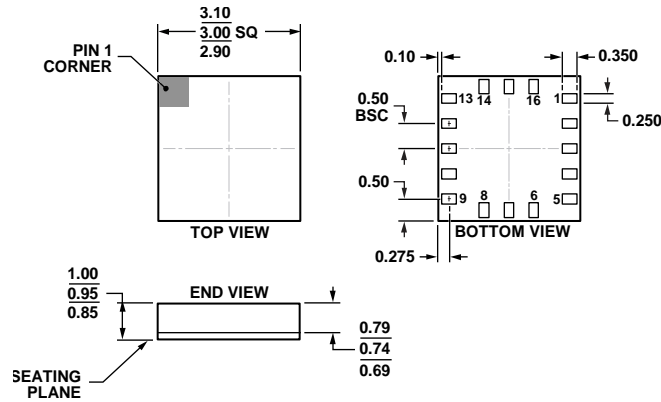
Table 27. Recommended Soldering Profile^{1, 2}

Profile Feature	Condition	
	Sn63/Pb37	Pb-Free
Average Ramp Rate from Liquid Temperature (T_L) to Peak Temperature (T_P)	3°C/sec max	3°C/sec max
Preheat		
Minimum Temperature (T_{SMIN})	100°C	150°C
Maximum Temperature (T_{SMAX})	150°C	200°C
Time from T_{SMIN} to T_{SMAX} (t_s)	60 sec to 120 sec	60 sec to 180 sec
T_{SMAX} to T_L Ramp-Up Rate	3°C/sec max	3°C/sec max
Liquid Temperature (T_L)	183°C	217°C
Time Maintained Above T_L (t_L)	60 sec to 150 sec	60 sec to 150 sec
Peak Temperature (T_P)	240 + 0/-5°C	260 + 0/-5°C
Time of Actual T_P - 5°C (t_p)	10 sec to 30 sec	20 sec to 40 sec
Ramp-Down Rate	6°C/sec max	6°C/sec max
Time 25°C to Peak Temperature	6 minutes max	8 minutes max

¹ Based on JEDEC Standard J-STD-020D.1.

² For best results, the soldering profile should be in accordance with the recommendations of the manufacturer of the solder paste used.

OUTLINE DIMENSIONS



07-30-2012-B

Figure 64. 16-Terminal Land Grid Array [LGA]
(CC-16-3)
Solder Terminations Finish Is Au over Ni
Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Measurement Range (g)	Specified Voltage (V)	Temperature Range	Package Description	Package Option	Branding Code
ADXL346ACCZ-R2	±2, ±4, ±8, ±16	2.6	-40°C to +85°C	16-Terminal Land Grid Array [LGA]	CC-16-3	Y2Z
ADXL346ACCZ-RL	±2, ±4, ±8, ±16	2.6	-40°C to +85°C	16-Terminal Land Grid Array [LGA]	CC-16-3	Y2Z
ADXL346ACCZ-RL7	±2, ±4, ±8, ±16	2.6	-40°C to +85°C	16-Terminal Land Grid Array [LGA]	CC-16-3	Y2Z
EVAL-ADXL346Z				Evaluation Board		
EVAL-ADXL346Z-DB				Inertial Sensor Development/Data Logger Board		
EVAL-ADXL346Z-M				Analog Devices Inertial Sensor Evaluation System, Includes ADXL346 Satellite		
EVAL-ADXL346Z-S				ADXL346 Satellite, Standalone		

¹ Z = RoHS Compliant Part.

NOTES

I²C refers to a communications protocol originally developed by Philips Semiconductors (now NXP Semiconductors).

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